

DATA SHEET



TEF6700HL

Car radio tuner front-end
for digital IF

Product specification

2003 Oct 21

Car radio tuner front-end for digital IF

TEF6700HL

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1 FEATURES

- FM mixer for conversion of FM RF (64 to 108 MHz and US weather band) to IF of 10.7 MHz; the mixer provides inherent image rejection and can be switched from 'low' injection to 'high' injection Local Oscillator (LO) via the I²C-bus; two different mixer conversion gains can be selected via the I²C-bus
- AGC PIN diode drive circuit for FM RF AGC; AGC detection at FM mixer input; the AGC PIN diode drive can be activated via the I²C-bus as a local function for search tuning; AGC threshold is a programmable and keyed function switchable via the I²C-bus
- Digital alignment circuit for bus controlled matching of oscillator tuning voltage to FM antenna tank circuit tuning voltage
- Buffer output for weather band flag
- FM IF linear amplifier with high dynamic input range; amplifier gain can be switched to two different positions via the I²C-bus
- FM IF AGC with multiplexed inputs for FM and In-Band On-Channel Digital Audio Broadcast (IBOC DAB)
- Buffer amplifier for FM IBOC DAB IF bypassing second IF filter can be activated via I²C-bus
- AM mixer 1 for conversion of AM RF to AM IF1 10.7 MHz
- AM RF PIN diode drive circuit and RF JFET conductance control by AGC cascode drive circuit; AGC threshold detection at AM mixer 1 and IF2 AGC input; threshold is programmable via the I²C-bus
- AM noise blanker with blanking at AM IF2; AM noise blanker can be deactivated via the I²C-bus
- AM mixer 2 for conversion of AM IF1 to AM IF2 450 kHz
- AM IF2 AGC
- Multiplexer to output AM IF2 AGC and FM IF AGC to a car IF DSP IC (e.g. SAA7724H)
- AM/FM RF AGC monitor output
- Level detector for AM and FM with temperature compensated output voltage; starting point and slope of level output is programmable via the I²C-bus



- IF counter for AM IF2 and FM IF
- LC tuner oscillator providing mixer frequencies for FM mixer and AM mixer 1
- Crystal oscillator providing mixer frequencies for AM mixer 2 and reference for synthesizer PLL, IF count and timing for Radio Data System (RDS) update
- Fast synthesizer PLL tuning system with local control for inaudible RDS updating
- Timing function for RDS update algorithm and control signal output for a car IF DSP IC (e.g. SAA7724H)
- Three hardware programmable I²C-bus addresses; pin BUSENABLE; two software controlled flag outputs
- Several test modes for fast IC and system tests.

2 GENERAL DESCRIPTION

The TEF6700HL is a single chip car radio tuner for AM, FM standard, FM IBOC DAB and weather band providing gain controlled output of FM IF and AM IF2 for a car IF DSP IC (e.g. SAA7724H) including the following functions:

- AM double conversion tuner for LW/MW/SW (31 m, 41 m and 49 m bands) with IF1 = 10.7 MHz and IF2 = 450 kHz
- FM single conversion tuner to an IF frequency of 10.7 MHz with integrated image rejection for US FM, Europe FM, Japan FM, East Europe FM and weather band reception; all bands can be selected using high side or low side Local Oscillator (LO) injection
- Tuning system includes VCO, crystal oscillator and PLL synthesizer on one chip.

3 ORDERING INFORMATION

TYPE NUMBER	PACKAGE		
	NAME	DESCRIPTION	VERSION
TEF6700HL	LQFP64	plastic low profile quad flat package; 64 leads; body 10 × 10 × 1.4 mm	SOT314-2

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4 QUICK REFERENCE DATA

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
V _{DDA(n)}	analog supply voltages 1 to 4 and 6		8	8.5	9	V
V _{DDA5}	analog supply voltage 5		4.75	5	5.25	V
V _{DDD}	digital supply voltage		4.75	5	5.25	V
I _{DDA(n)}	sum of analog supply currents 1 to 4 and 6	FM Japan mode	42	55	71	mA
		AM mode	32	43	56	mA
I _{DDA5}	analog supply current 5	FM Japan mode	7.5	10	12	mA
		AM mode	14	18.5	23	mA
I _{DDD}	digital supply current	FM Japan mode	25	32	39.5	mA
		AM mode	21	26	32	mA
f _{AM(ant)}	AM input frequency	LW	0.144	–	0.288	MHz
		MW	0.522	–	1.710	MHz
		SW	5.73	–	9.99	MHz
f _{FM(ant)}	FM input frequency		64	–	108	MHz
f _{FM(WB)(ant)}	FM weather band input frequency		162.4	–	162.55	MHz
T _{amb}	ambient temperature		–40	–	+85	°C
AM overall system parameters; see Fig.10						
g _{m(conv)}	AM mixer 1 conversion transconductance $\frac{I_{IF1}}{V_{RF}}$		3.9	5.2	6.5	$\frac{mA}{V}$
F _{AMMIX1}	noise figure of AM mixer 1		–	4.5	7.1	dB
IP3	3rd-order input intercept point	R _L = 2.6 kΩ (AC load between output pins); Δf = 300 kHz	135	138	–	dBμV
g _{m(conv)}	AM mixer 2 conversion transconductance $\frac{I_{IF2}}{V_{IF1}}$		3.2	4.3	5.4	$\frac{mA}{V}$
ΔAGC	AM IF2 amplifier gain control range $\frac{V_{IFstop}}{V_{IFstart}}$		52	–	–	dB
FM overall system parameters; see Fig.10						
FM MIXER						
g _{m(conv)}	conversion transconductance $\frac{I_{IF}}{V_{RF}}$	MIXGAIN = 0	8.5	12.5	18	$\frac{mA}{V}$
		MIXGAIN = 1	17	25	36	$\frac{mA}{V}$
F _{FMmixer}	noise figure of FM mixer	MIXGAIN = 0	–	3.5	4.6	dB
		MIXGAIN = 1	–	2.4	–	dB

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
IP3	3rd-order input intercept point	MIXGAIN = 0	113	117	–	dB μ V
		MIXGAIN = 1	–	108	–	dB μ V
FM AMPLIFIER						
G _{IFAMP}	FM IF amplifier gain $\frac{V_{FMIFAMP\text{OUT}}}{V_{IFAMPIN-IFAMPDEC}}$	R _L = 330 Ω ; V _{IFAMPIN-IFAMPDEC} = 1 mV IFGAIN = 0	–	10.5	–	dB
		IFGAIN = 1	–	14	–	dB
G _{IBOCAMP}	FM IBOC amplifier gain	IBOC = 1	–	8	–	dB
IP3 _{IF}	3rd-order input intercept point	IFGAIN = 0	–	123	–	dB μ V
		IFGAIN = 1	–	119	–	dB μ V
IP3 _{IBOC}	3rd-order input intercept point IBOC mode	IBOC = 1	–	123	–	dB μ V
F _{IFAMP}	noise figure of FM IF amplifier	IFGAIN = 0	–	10	–	dB
		IFGAIN = 1	–	8.3	–	dB
F _{IBOCAMP}	noise figure of FM IBOC amplifier	IBOC = 1	–	7	–	dB
Δ AGC	FM IF amplifier gain control range $\frac{V_{IF\text{stop}}}{V_{IF\text{start}}}$		59	–	–	dB

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5 BLOCK DIAGRAM

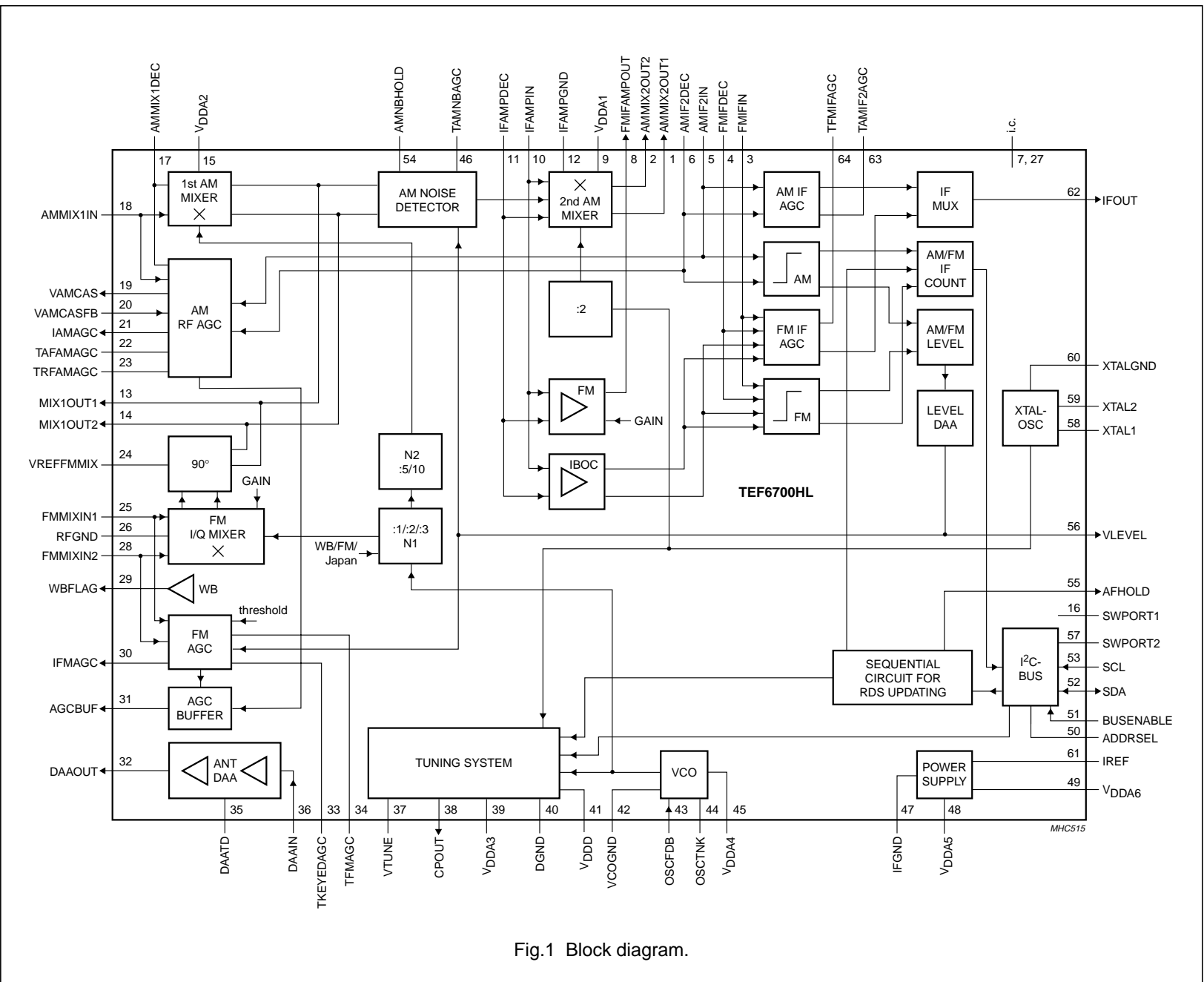


Fig.1 Block diagram.

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6 PINNING

SYMBOL	PIN	DESCRIPTION
AMMIX2OUT1	1	2nd AM mixer output 1 (450 kHz)
AMMIX2OUT2	2	2nd AM mixer output 2 (450 kHz)
FMIFIN	3	FM IF AGC and FM level detector input
FMIFDEC	4	FM IF AGC and FM level detector decoupling
AMIF2IN	5	AM IF2 AGC and AM level detector input (450 kHz)
AMIF2DEC	6	AM IF2 AGC and AM level detector decoupling
i.c.	7	internal connected
FMIFAMPOUT	8	FM IF amplifier output (10.7 MHz)
V _{DDA1}	9	analog supply voltage 1 (8.5 V) for FM IF amplifier
IFAMPIN	10	FM IF amplifier and AM mixer 2 input (10.7 MHz)
IFAMPDEC	11	FM IF amplifier and AM mixer 2 decoupling
IFAMPGND	12	ground for FM IF amplifier
MIX1OUT1	13	FM mixer and AM mixer 1 IF output 1 (10.7 MHz)
MIX1OUT2	14	FM mixer and AM mixer 1 IF output 2 (10.7 MHz)
V _{DDA2}	15	analog supply voltage 2 (8.5 V) for FM and AM RF
SWPORT1	16	software programmable port 1
AMMIX1DEC	17	AM mixer 1 decoupling
AMMIX1IN	18	AM mixer 1 input
VAMCAS	19	output for AM RF cascode AGC
VAMCASFB	20	feedback input for AM RF cascode AGC
IAMAGC	21	PIN diode drive current output of AM front-end AGC
TAFAMAGC	22	AF time constant of AM front-end AGC
TRFAMAGC	23	RF time constant of AM front-end AGC
VREFFMMIX	24	reference voltage for FM RF mixer
FMMIXIN1	25	FM mixer input 1
RFGND	26	RF ground
i.c.	27	internal connected
FMMIXIN2	28	FM mixer input 2
WBFLAG	29	buffered weather band flag output
IFMAGC	30	PIN diode drive current output of FM front-end AGC
AGCBUF	31	monitor current output of FM and AM front-end AGC
DAAOUT	32	output of digital auto alignment circuit for antenna tank circuit
TKEYEDAGC	33	time constant of keyed FM front-end AGC
TFMAGC	34	time constant of FM front-end AGC
DAATD	35	temperature compensation diode of digital auto alignment circuit for antenna tank circuit
DAAIN	36	input of digital auto alignment circuit for antenna tank circuit
VTUNE	37	VCO tuning voltage
CPOUT	38	charge pump output
V _{DDA3}	39	analog supply voltage 3 (8.5 V) for tuning PLL
DGND	40	digital ground

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SYMBOL	PIN	DESCRIPTION
V _{DDD}	41	digital supply voltage (5 V)
VCOGND	42	VCO ground
OSCFDB	43	VCO feedback input
OSCTNK	44	VCO tank circuit
V _{DDA4}	45	analog supply voltage 4 (8.5 V) for VCO
TAMNBAGC	46	AGC time constant for AM IF noise blanker
IFGND	47	IF AGC ground
V _{DDA5}	48	analog supply voltage 5 (5 V) for on-chip power supply
V _{DDA6}	49	analog supply voltage 6 (8.5 V) for on-chip power supply
ADDRSEL	50	hardware address select for I ² C-bus
BUSENABLE	51	enable input for I ² C-bus
SDA	52	I ² C-bus data line input and output
SCL	53	I ² C-bus clock line input
AMNBHOLD	54	AM noise blanker threshold
AFHOLD	55	AF hold flag output for a car IF DSP IC (e.g. SAA7724H)
VLEVEL	56	level voltage output for AM and FM
SWPORT2	57	software programmable port 2
XTAL1	58	crystal oscillator 1
XTAL2	59	crystal oscillator 2
XTALGND	60	crystal oscillator ground
IREF	61	reference current for power supply
IFOUT	62	multiplexer output for FM IF AGC and AM IF2 AGC
TAMIF2AGC	63	time constant for AM IF2 AGC
TFMIFAGC	64	time constant for FM IF AGC

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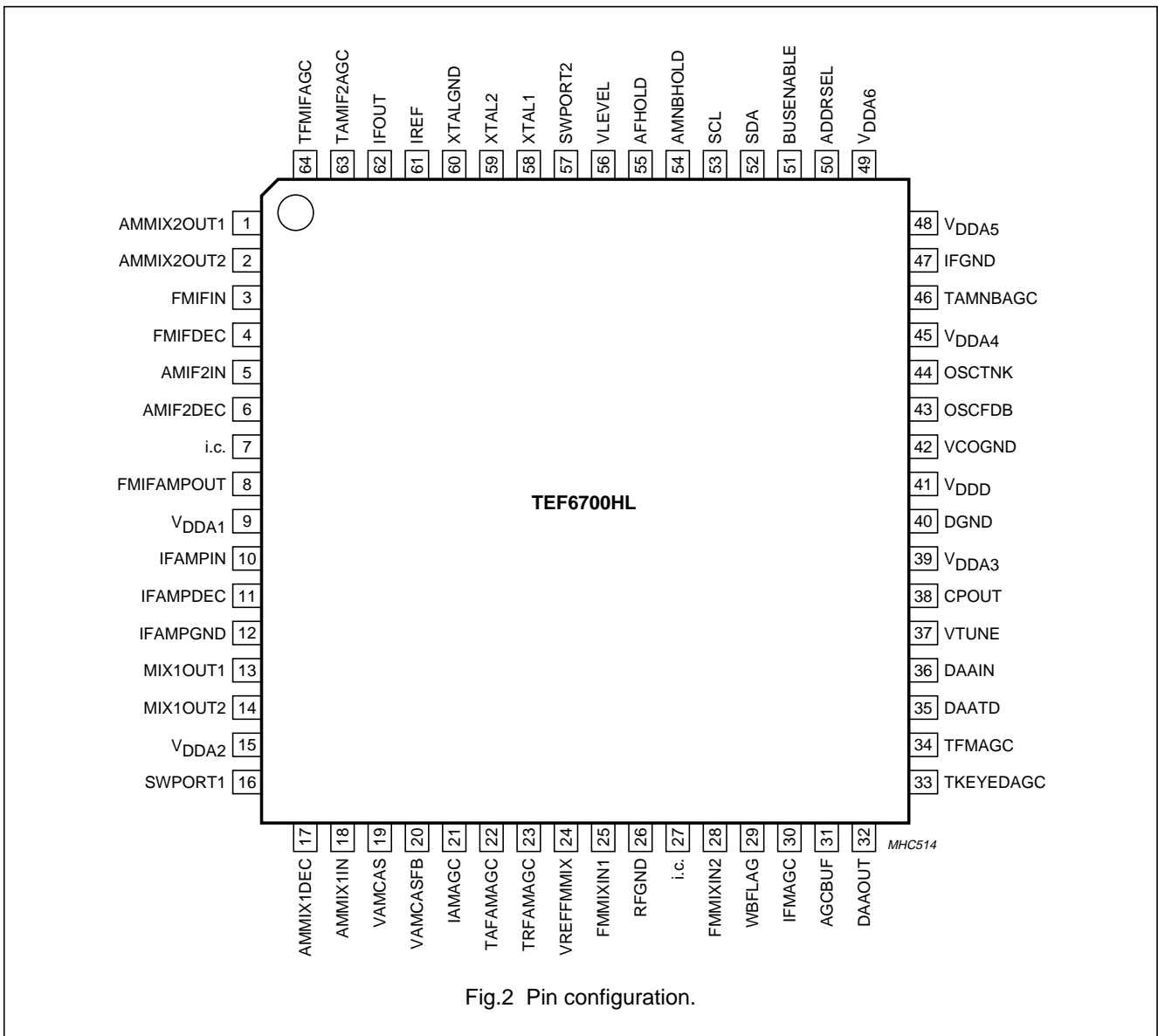


Fig.2 Pin configuration.

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7 FUNCTIONAL DESCRIPTION

7.1 FM I/Q-mixer

The FM quadrature mixer converts FM RF (64 to 162.55 MHz) to an IF of 10.7 MHz. The FM mixer provides inherent image rejection and high RF sensitivity. The image rejection can be switched from 'low' injection Local Oscillator (LO) to 'high' injection LO via the I²C-bus. The mixer gain can be increased by 6 dB via the I²C-bus. In this case the threshold of the FM keyed AGC has to be lowered by 6 dB to prevent the mixer from being overloaded.

The mixer is capable of tuning the:

- US FM and US IBOC DAB from 87.9 to 107.9 MHz
- US weather FM from 162.4 to 162.55 MHz
- Europe FM from 87.5 to 108 MHz
- Japan FM from 76 to 91 MHz
- East Europe FM from 64 to 74 MHz.

7.2 Buffer output for weather band flag

The buffer output (pin WBFLAG) is HIGH for weather band mode.

7.3 VCO

The varactor tuned LC oscillator provides the local oscillator signal for both FM and AM mixer 1. It has a frequency range from 159.9 to 248.2 MHz.

7.4 Crystal oscillator

The crystal oscillator provides a 20.5 MHz signal that is used for:

- Reference frequency for frequency synthesizer PLL
- Local oscillator for AM mixer 2
- Reference frequency for the IF counter
- Timing signal for the RDS update algorithm.

7.5 PLL

The fast synthesizer PLL tuning system with local control is used for inaudible RDS updating.

7.6 DAA

To reduce the number of manual alignments in production the following I²C-bus controlled Digital Auto Alignment (DAA) functions are included:

- FM RF DAA
 - 7-bit DAA circuitry for the conversion of the VCO tuning voltage to a controlled alignment voltage for the FM antenna tank circuit
- FM and AM level DAA
 - Level DAA circuitry for alignment of slope (3-bit) and starting point (5-bit) of the level curve.

7.7 FM keyed AGC

The AGC threshold is programmable and the keyed AGC function is switchable via the I²C-bus. AGC detection occurs at the input of the FM mixer. If the keyed AGC function is activated, the AGC is keyed only by the narrow-band level. The AGC PIN diode drive can be activated via the I²C-bus as a local function for search tuning. The AGC sources a constant 10 mA current into the FM PIN diode in AM mode.

7.8 FM IF amplifier

The FM IF amplifier provides 10.5 or 14 dB gain, selectable via the I²C-bus with high linearity over a wide dynamic range. In IBOC DAB mode (activated via I²C-bus) FM IF is buffered and directly fed to FM IF AGC bypassing the second FM ceramic filter.

7.9 FM IF AGC

This is a linear FM IF AGC amplifier with multiplexed inputs for FM standard and FM IBOC DAB. The gain controlled FM IF is output by the multiplexed FM IF/AM IF2 buffer for a car IF DSP IC (e.g. SAA7724H).

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7.10 AM tuner including mixer 1 and mixer 2

The AM tuner is realized in a double conversion technique and is capable of selecting LW, MW and SW bands.

AM mixer 1 converts AM RF to an IF1 of 10.7 MHz, while AM mixer 2 converts IF1 of 10.7 MHz to IF2 of 450 kHz:

- LW from 144 to 288 kHz
- MW from 522 to 1710 kHz (US AM band)
- SW from 5.73 to 9.99 MHz (including the 31 m, 41 m and 49 m bands).

7.11 AM RF AGC

The AM wideband AGC in front of the AM mixer 1 is realized first by a cascoded NPN transistor, which controls the transconductance of the RF amplifier JFET with 10 dB of AGC range. Second, an AM PIN diode stage with 30 dB of AGC range is available. The minimum JFET drain source voltage is controlled by a DC feedback loop (pin VAMCASFB) in order to limit the cascode AGC range to 10 dB. If the cascode AGC is not required, a simple RF AGC loop is possible by using only a PIN diode. In some conditions, noise behaviour will increase. In this case pins VAMCAS and VAMCASFB have to be left open-circuit. In FM mode, the cascode switches off the JFET bias current to reduce total power consumption. The PIN diode is biased by 1 mA in FM mode.

The AGC detection points for AM RF AGC are at the AM mixer 1 input (threshold programmable via the I²C-bus) and the IF2 AGC input (fixed threshold).

7.12 AM noise blanker

The detection point for the AM noise blanker is the output stage of AM mixer 1, while blanking is realized at the output of mixer 2.

7.13 AM IF2 AGC

This is a linear AM IF2 AGC amplifier. The gain controlled AM IF2 is output via the multiplexed FM IF/AM IF2 buffer to a car IF DSP IC (e.g. SAA7724H).

7.14 FM IF and AM IF2 buffer

This buffers and multiplexes AM IF2 and FM IF to pin IFOUT.

7.15 FM and AM level detector

The FM and AM level detectors provide the temperature compensated output voltage. The starting points and slopes of the level detector outputs are programmable via the I²C-bus.

7.16 FM/AM RF AGC buffer

The output (open-collector) sinks a current which in AM mode is proportional to the voltage at pin TRFAMAGC and in FM mode proportional to the RF level detector voltage (pin TFMAGC) inside the FM AGC.

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8 LIMITING VALUES

In accordance with the Absolute Maximum Rating System (IEC 60134).

SYMBOL	PARAMETER	CONDITIONS	MIN.	MAX.	UNIT
V _{DDA1}	analog supply voltage 1 for FM IF amplifier	note 1	-0.3	+10	V
V _{DDA2}	analog supply voltage 2 for FM and AM RF	note 1	-0.3	+10	V
V _{DDA3}	analog supply voltage 3 for tuning PLL	note 1	-0.3	+10	V
V _{DDA4}	analog supply voltage 4 for VCO	note 1	-0.3	+10	V
V _{DDA5}	analog supply voltage 5 for on-chip power supply		-0.3	+6.5	V
V _{DDA6}	analog supply voltage 6 for on-chip power supply	note 1	-0.3	+10	V
V _{DDD}	digital supply voltage		-0.3	+6.5	V
T _{stg}	storage temperature		-55	+150	°C
T _{amb}	ambient temperature		-40	+85	°C
V _{es}	electrostatic handling voltage	note 2	-200	+200	V
		note 3	-2000	+2000	V

Notes

- To avoid damages and wrong operation it is necessary to keep all 8.5 V supply voltages at a higher level than any 5 V supply voltage. This is also necessary during power-on and power-down sequences. Precautions have to be provided in such a way that interferences can not pull down the 8.5 V supply below the 5 V supply.
- Machine model (R = 0 Ω, C = 200 pF).
- Human body model (R = 1.5 kΩ, C = 100 pF).

9 THERMAL CHARACTERISTICS

SYMBOL	PARAMETER	CONDITIONS	VALUE	UNIT
R _{th(j-a)}	thermal resistance from junction to ambient	in free air	58	K/W

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10 DC CHARACTERISTICS

$V_{DDA1} = V_{DDA2} = V_{DDA3} = V_{DDA4} = V_{DDA6} = 8.5\text{ V}$; $V_{DDA5} = 5\text{ V}$; $V_{DDD} = 5\text{ V}$; $T_{\text{amb}} = 25\text{ }^{\circ}\text{C}$; unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Supply voltage						
$V_{DDA(n)}$	analog supply voltages 1 to 4 and 6		8	8.5	9	V
V_{DDA5}	analog supply voltage 5		4.75	5	5.25	V
V_{DDD}	digital supply voltage		4.75	5	5.25	V
Supply current in FM mode						
I_{DDA1}	analog supply current 1 for FM IF amplifier	IBOC mode	–	7.9	–	mA
		no IBOC mode	–	8.9	–	mA
I_{DDA2}	analog supply current 2 for RF		4.7	5.7	6.7	mA
I_{DDA3}	analog supply current 3 for tuning PLL		–	4	–	mA
I_{DDA4}	analog supply current 4 for VCO		5.2	6.5	7.8	mA
I_{DDA5}	analog supply current 5 for on-chip power supply	Europe/US band	–	6.4	–	mA
		Japan/East Europe band	7.5	10	12	mA
I_{DDA6}	analog supply current 6 for on-chip power supply		–	20.5	–	mA
I_{DDD}	digital supply current	Europe/US band	20	25	31	mA
		Japan/East Europe band	25	32	39.5	mA
$I_{MIX1OUT1}$	bias current of FM mixer output 1		4.8	6	7.2	mA
$I_{MIX1OUT2}$	bias current of FM mixer output 2		4.8	6	7.2	mA
Supply current in AM mode						
I_{DDA1}	analog supply current 1 for AM mixer 2		–	140	–	μA
I_{DDA2}	analog supply current 2 for RF		–	2.6	–	mA
I_{DDA3}	analog supply current 3 for tuning PLL		–	2.2	–	mA
I_{DDA4}	analog supply current 4 for VCO		5	6.5	8	mA
I_{DDA5}	analog supply current 5 for on-chip power supply		14	18.5	23	mA
I_{DDA6}	analog supply current 6 for on-chip power supply		–	16.7	–	mA
I_{DDD}	digital supply current		21	26	32	mA
$I_{MIX1OUT1}$	bias current of AM mixer 1 output 1		4.8	6	7.2	mA
$I_{MIX1OUT2}$	bias current of AM mixer 1 output 2		4.8	6	7.2	mA
$I_{AMMIX2OUT1}$	bias current of AM mixer 2 output 1		3.6	4.5	5.4	mA
$I_{AMMIX2OUT2}$	bias current of AM mixer 2 output 2		3.6	4.5	5.4	mA
On-chip power supply reference current generator: pin IREF						
$V_{O(\text{ref})}$	output reference voltage		4	4.25	4.5	V
R_O	output resistance		–	10	–	$\text{k}\Omega$
$I_{O(\text{source})(\text{max})}$	maximum output source current		–100	–	+100	nA

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11 AC CHARACTERISTICS

$V_{\text{AMMIX2OUT1}} = V_{\text{AMMIX2OUT2}} = V_{\text{MIX1OUT1}} = V_{\text{MIX1OUT2}} = V_{\text{DDA1}} = V_{\text{DDA2}} = V_{\text{DDA3}} = V_{\text{DDA4}} = V_{\text{DDA6}} = 8.5 \text{ V}$; $V_{\text{DDA5}} = 5 \text{ V}$;
 $V_{\text{DD}} = 5 \text{ V}$; $T_{\text{amb}} = 25 \text{ }^\circ\text{C}$; see Fig.10; all AC values are given in RMS; unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Voltage controlled oscillator						
f_{osc}	oscillator frequency		159.9	–	248.2	MHz
C/N	carrier-to-noise ratio	$f_{\text{osc}} = 200 \text{ MHz}$; $\Delta f = 10 \text{ kHz}$; $B = 1 \text{ Hz}$	–	97	–	dBc
RR	ripple rejection $\frac{\Delta f_{\text{osc}}}{f_{\text{osc}}}$	$f_{\text{ripple}} = 100 \text{ Hz}$; $V_{\text{DDA4(ripple)}} = 100 \text{ mV}$; $f_{\text{osc}} = 200 \text{ MHz}$	92	99	–	dB
Crystal oscillator						
f_{xtal}	crystal frequency		–	20.5	–	MHz
C/N	carrier-to-noise ratio	$f_{\text{xtal}} = 20.5 \text{ MHz}$; $\Delta f = 10 \text{ kHz}$	–	112	–	$\frac{\text{dBc}}{\sqrt{\text{Hz}}}$
CIRCUIT INPUTS: PINS XTAL1, XTAL2 AND XTALGND; note 1						
V_{xtal}	crystal voltage		80	100	160	mV
V_{XTAL1} , V_{XTAL2}	DC bias voltage		1.7	2.1	2.5	V
R_i	real part of input impedance	$V_{\text{XTAL1}} - V_{\text{XTAL2}} = 1 \text{ mV}$	–250	–	–	Ω
C_i	input capacitance		8	10	12	pF
Synthesizer						
PROGRAMMABLE DIVIDER						
N_{prog}	programmable divider ratio		512	–	32767	
ΔN_{step}	programmable divider step size		–	1	–	
CHARGE PUMP: PIN CPOUT						
$I_{\text{sink(cp1)l}}$	low charge pump 1 peak sink current	FM weather band mode; $0.4 \text{ V} < V_{\text{CPOUT}} < 7.6 \text{ V}$; $f_{\text{VCO}} > f_{\text{ref}} \times N_{\text{prog}}$	–	300	–	μA
$I_{\text{source(cp1)l}}$	low charge pump 1 peak source current	FM weather band mode; $0.4 \text{ V} < V_{\text{CPOUT}} < 7.6 \text{ V}$; $f_{\text{VCO}} < f_{\text{ref}} \times N_{\text{prog}}$	–	–300	–	μA
$I_{\text{sink(cp1)h}}$	high charge pump 1 peak sink current	$0.4 \text{ V} < V_{\text{CPOUT}} < 7.6 \text{ V}$; $f_{\text{VCO}} > f_{\text{ref}} \times N_{\text{prog}}$ AM mode	–	1	–	mA
		FM East Europe band	–	1	–	mA
$I_{\text{source(cp1)h}}$	high charge pump 1 peak source current	$0.4 \text{ V} < V_{\text{CPOUT}} < 7.6 \text{ V}$; $f_{\text{VCO}} < f_{\text{ref}} \times N_{\text{prog}}$ AM mode	–	–1	–	mA
		FM East Europe band	–	–1	–	mA

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
$I_{\text{sink(cp2)}}$	charge pump 2 peak sink current	FM standard or FM Japan mode; $f_{\text{VCO}} > f_{\text{ref}} \times N_{\text{prog}}$; $0.3 \text{ V} < V_{\text{CPOUT}} < 7.1 \text{ V}$	–	130	–	μA
$I_{\text{source(cp2)}}$	charge pump 2 peak source current	FM standard or FM Japan mode; $f_{\text{VCO}} < f_{\text{ref}} \times N_{\text{prog}}$; $0.3 \text{ V} < V_{\text{CPOUT}} < 7.1 \text{ V}$	–	–130	–	μA
CHARGE PUMP: PIN VTUNE						
$I_{\text{sink(cp3)}}$	charge pump 3 peak sink current	FM standard or FM Japan mode; $f_{\text{VCO}} > f_{\text{ref}} \times N_{\text{prog}}$; $0.4 \text{ V} < V_{\text{VTUNE}} < 7.6 \text{ V}$	–	3	–	mA
$I_{\text{source(cp3)}}$	charge pump 3 peak source current	FM standard or FM Japan mode; $f_{\text{VCO}} < f_{\text{ref}} \times N_{\text{prog}}$; $0.4 \text{ V} < V_{\text{VTUNE}} < 7.6 \text{ V}$	–	–3	–	mA
Antenna Digital Auto Alignment (DAA)						
DAA INPUT: PIN DAAIN						
$I_{\text{bias(cp)}}$	charge pump buffer input bias current	$V_{\text{DAAIN}} = 0.4 \text{ to } 8 \text{ V}$	–10	–	+10	nA
$V_{\text{i(cp)}}$	charge pump buffer input voltage		0	–	8.5	V
DAA OUTPUT: PIN DAAOUT						
$V_{\text{o(AM)}}$	DAA output voltage in AM mode	$I_{\text{DAAOUT}} < 100 \mu\text{A}$	–	–	0.3	V
$V_{\text{o(FM)}}$	DAA output voltage in FM mode	$V_{\text{DAATD}} = 0.45 \text{ V}$ minimum value; data byte 2 = 10000000 (n = 0); $V_{\text{DAAIN}} = 0.5 \text{ V}$ data byte 2 = 10101010 (n = 42); $V_{\text{DAAIN}} = 2 \text{ V}$ data byte 2 = 11010101 (n = 85); $V_{\text{DAAIN}} = 2 \text{ V}$; note 2 data byte 2 = 10000000 (n = 0); $V_{\text{DAAIN}} = 4 \text{ V}$; note 2 data byte 2 = 11000000 (n = 64); $V_{\text{DAAIN}} = 4 \text{ V}$ maximum value; data byte 2 = 11111111 (n = 127); $V_{\text{DAAIN}} = 4.7 \text{ V}$	–	–	0.5	V
			1.2	1.4	1.6	V
			2.3	2.6	2.9	V
			0.4	0.65	1	V
			3.8	4	4.2	V
			8	–	8.5	V
$V_{\text{o(n)}}$	DAA output noise voltage	data byte 2 = 11000000 (n = 64); FM mode; $V_{\text{DAAIN}} = 4 \text{ V}$; $V_{\text{DAATD}} = 0.45 \text{ V}$; B = 300 Hz to 22 kHz	–	30	100	μV
$\Delta V_{\text{o(T)}}$	DAA output voltage variation with temperature	$T_{\text{amb}} = -40 \text{ to } +85 \text{ }^\circ\text{C}$; data byte 2 = 11000000 (n = 64)	–8	–	+8	mV

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
$\Delta V_{o(\text{step})}$	DAA step accuracy	FM mode; $n = 0$ to 127; $V_{\text{DAAOUT}} = 0.5$ to 8 V; $V_{\text{DAAIN}} = 2$ V; $V_{\text{DAATD}} = 0.45$ V; note 3	$0.5V_{\text{LSB}}$	V_{LSB}	$1.5V_{\text{LSB}}$	mV
$\Delta V_{o(\text{sink})}$	DAA output voltage variation caused by sink current	$V_{\text{DAAIN}} = 4$ V; $I_{\text{DAAOUT}} = 50$ μA ; note 3	$-V_{\text{LSB}}$	–	$+V_{\text{LSB}}$	mV
$\Delta V_{o(\text{source})}$	DAA output voltage variation caused by source current	$V_{\text{DAAIN}} = 4$ V; $I_{\text{DAAOUT}} = -50$ μA ; note 3	$-V_{\text{LSB}}$	–	$+V_{\text{LSB}}$	mV
t_{st}	DAA output settling time	$V_{\text{DAAOUT}} = 0.2$ to 8.25 V; $C_{\text{L}} = 270$ pF	–	20	30	μs
RR	ripple rejection $\frac{V_{\text{DAAOUT}}}{V_{\text{DDA3}}}$	data byte 2 = 10101011 ($n = 43$); FM mode; $V_{\text{DAAIN}} = 4$ V; $V_{\text{DAATD}} = 0.45$ V; $f_{\text{ripple}} = 100$ Hz; $V_{\text{DDA3(ripple)}} = 100$ mV	–	65	–	dB
C_{L}	DAA output load capacitance		–	–	270	pF
DAA TEMPERATURE COMPENSATION: PIN DAATD						
I_{source}	compensation diode source current	$V_{\text{DAATD}} = 0.2$ to 1.2 V	–50	–40	–30	μA
TC_{source}	temperature coefficient of compensation diode source current	$V_{\text{DAATD}} = 0.2$ to 1.2 V; $T_{\text{amb}} = -40$ to $+85$ °C	–300	–	+300	$\frac{10^{-6}}{\text{K}}$
IF counter (FM IF or AM IF2 counter)						
N_{IF}	IF counter length for AM and FM		–	8	–	bit
AM MODE						
V_{sen}	sensitivity voltage (pin AMIF2IN to AMIF2DEC)	$m = 0$	–	30	70	μV
N	counter result (decimal)	$V_{\text{AMIF2IN-AMIF2DEC}} = 200$ μV period = 2 ms period = 20 ms	– –	132 40	– –	

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FM MODE						
V_{sen}	sensitivity voltage (pin FMIFIN to FMIFDEC)		–	60	100	μV
N	counter result (decimal)	period = 2 ms; $V_{\text{FMIFIN-FMIFDEC}} = 100 \mu\text{V}$ prescaler ratio = 10	–	92	–	
		prescaler ratio = 100	–	214	–	
		period = 20 ms; $V_{\text{FMIFIN-FMIFDEC}} = 100 \mu\text{V}$ prescaler ratio = 10	–	152	–	
		prescaler ratio = 100	–	92	–	
I²C-bus address select: pin ADDRSEL						
R_L	load resistance to ground	1st I ² C-bus address	1	–	–	$\text{M}\Omega$
		2nd I ² C-bus address	108	120	132	$\text{k}\Omega$
		3rd I ² C-bus address	29.7	33	36.3	$\text{k}\Omega$
I²C-bus enable: pin BUENABLE						
V_{IL}	LOW-level input voltage		–0.3	–	+1	V
V_{IH}	HIGH-level input voltage		2	–	$V_{\text{DDD}} + 0.3$	V
Software programmable ports: pins SWPORT1 and SWPORT2						
$I_{\text{sink(max)}}$	maximum sink current	SWPORT1 = 1	1	–	1.6	mA
		SWPORT2 = 1	1	–	1.6	mA
Weather band flag: pin WBFLAG						
$I_{\text{source(max)}}$	maximum source current	$R = 560 \Omega$	–6	–5	–4	mA
$R_{\text{i(shunt)}}$	internal shunt resistance to ground		–	50	–	$\text{k}\Omega$
$V_{\text{o(FM)(max)}}$	maximum output voltage for FM mode	measured with respect to pin RFGND	0	–	0.2	V
$V_{\text{o(WB)}}$	output voltage for weather band mode	measured with respect to pin RFGND	4	–	5	V
AM signal channel						
AM RF AGC: PINS AMMIX1IN AND AMMIX1DEC						
$V_{\text{i(RF)(p)}}$	RF input voltage for wideband AGC start level (peak value)	$m = 0.3; f_{\text{AF}} = 1 \text{ kHz}$				
		AGC[1:0] = 00	360	520	735	mV
		AGC[1:0] = 01	480	680	960	mV
		AGC[1:0] = 10	590	840	1190	mV
		AGC[1:0] = 11	690	980	1390	mV
AM IF AGC STAGE INPUTS: PINS AMIF2IN AND AMIF2DEC						
$V_{\text{i(IF2)}}$	IF2 input voltage	AGC start level				
		$m = 0$	0.09	0.12	0.19	V
		$m = 0.8$	0.07	0.10	0.14	V

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
AM RF AGC PIN DIODE DRIVE: PIN IAMAGC						
$I_{\text{sink(max)}}$	maximum AGC sink current	$V_{\text{IAMAGC}} = 2.8 \text{ V}$	9	14	18	mA
I_{sink}	AGC sink current	FM mode; $V_{\text{IAMAGC}} = 2.8 \text{ V}$	1	1.5	2	mA
R_o	output resistance	$I_{\text{IAMAGC}} = 1 \mu\text{A}$	0.5	–	–	MΩ
C_o	output capacitance		–	5	7	pF
AM RF AGC CASCODE STAGE: PIN VAMCAS						
V_{cas}	cascode voltage	$V_{\text{AMMIX1IN-AMMIX1DEC}}$ below threshold; maximum gain	4.5	5	5.5	V
I_{cas}	cascode transistor base current capability		100	–	–	μA
$I_{\text{cas(off)}}$	cascode transistor base off current	FM mode	–	–	100	nA
AM RF AGC CASCODE STAGE: PIN VAMCASFB						
$V_{\text{cas(FB)}}$	cascode voltage	$V_{\text{AMMIX1IN-AMMIX1DEC}}$ above threshold; minimum gain	0.2	0.28	0.35	V
$I_{\text{cas(FB)}}$	cascode feedback sense current		0	0.4	1	μA
AM RF AGC TRANSCONDUCTANCE BUFFER: PIN AGCBUF; note 4						
$g_{\text{m(buf)}}$	buffer transconductance $\frac{\Delta I_{\text{AGCBUF}}}{\Delta V_{\text{TRFAMAGC}}}$	AM mode; $\Delta V_{\text{TRFAMAGC}} = 50 \text{ mV to } 0.4 \text{ V}$	0.85	1.1	1.35	mS
$I_{\text{sink(max)}}$	maximum sink current	AM mode; open-collector; $\Delta V_{\text{TRFAMAGC}} = 0.8 \text{ V}$	450	500	560	μA
$I_{\text{source(max)}}$	maximum source current	AM mode; $\Delta V_{\text{TRFAMAGC}} < 50 \text{ mV}$	–	–	–30	μA
$V_{\text{o(n)}}$	buffer output noise voltage	AM mode; $V_{\text{DDA2}} - V_{\text{AGCBUF}} = 1 \text{ V}$ (voltage across external pull-up resistor); $B = 400 \text{ Hz to } 20 \text{ kHz}$	–	10	15	μV
AM MIXER 1 (IF1 = 10.7 MHz)						
<i>Mixer inputs: pins AMMIX1IN and AMMIX1DEC</i>						
R_i	input resistance		15	25	40	kΩ
C_i	input capacitance		2.5	5	7.5	pF
V_i	DC input voltage		2.3	2.7	3.1	V
$V_{i(\text{max})}$	maximum input voltage	1 dB compression point of AM mixer 1 output; $m = 0$	500	–	–	mV

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
<i>Mixer outputs: pins MIX1OUT1 and MIX1OUT2</i>						
R _o	output resistance		100	–	–	kΩ
C _o	output capacitance		–	4	7	pF
V _{o(max)(p-p)}	maximum output voltage (peak-to-peak value)		12	15	–	V
I _{bias}	mixer bias current	AM mode	4.8	6	7.2	mA
<i>Mixer</i>						
g _{m(conv)}	conversion transconductance $\frac{I_{IF1}}{V_{RF}}$		3.9	5.2	6.5	$\frac{\text{mA}}{\text{V}}$
g _{m(conv)(T)}	conversion transconductance variation with temperature $\frac{\Delta g_{m(conv)}}{g_{m(conv)} \times \Delta T}$		–	-9×10^{-4}	–	K ⁻¹
IP3	3rd-order intermodulation	R _L = 2.6 kΩ (AC load between output pins); Δf = 300 kHz	135	138	–	dBμV
IP2	2nd-order intermodulation	R _L = 2.6 kΩ (AC load between output pins)	–	170	–	dBμV
V _{i(n)(eq)}	equivalent input noise voltage	R _{gen} = 750 Ω; R _L = 2.6 kΩ (AC load between output pins)	–	5.8	8	$\frac{\text{nV}}{\sqrt{\text{Hz}}}$
F	noise figure of AM mixer 1		–	4.5	7.1	dB
AM MIXER 2 (IF2 = 450 kHz)						
<i>Mixer inputs: pins IFAMPIN and IFAMPDEC</i>						
R _i	input resistance		–	330	–	Ω
C _i	input capacitance		–	5	7	pF
V _I	DC input voltage		2.4	2.7	3	V
V _{i(max)(p)}	maximum input voltage (peak value)	1 dB compression point of AM mixer 2 output	1.1	1.4	–	V
<i>Mixer outputs: pins AMMIX2OUT1 and AMMIX2OUT2</i>						
R _o	output resistance		50	–	–	kΩ
C _o	output capacitance		–	4	7	pF
V _{o(max)(p-p)}	maximum output voltage (peak-to-peak value)		12	15	–	V
I _{bias}	mixer bias current	AM mode	3.6	4.5	5.4	mA
I _L	mixer leakage current	FM mode	–	–	10	μA

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
<i>Mixer</i>						
$g_{m(\text{conv})}$	conversion transconductance $\frac{I_{\text{IF2}}}{V_{\text{IF1}}}$		3.2	4.3	5.4	$\frac{\text{mA}}{\text{V}}$
$g_{m(\text{conv})(T)}$	conversion transconductance variation with temperature $\frac{\Delta g_{m(\text{conv})}}{g_{m(\text{conv})} \times \Delta T}$		–	-9×10^{-4}	–	K^{-1}
IP3	3rd-order intermodulation	$R_L = 1.5 \text{ k}\Omega$ (AC load between output pins); $\Delta f = 300 \text{ kHz}$	134	137	–	$\text{dB}\mu\text{V}$
IP2	2nd-order intermodulation	$R_L = 1.5 \text{ k}\Omega$ (AC load between output pins)	–	170	–	$\text{dB}\mu\text{V}$
$V_{i(n)(\text{eq})}$	equivalent input noise voltage	$R_{\text{gen}} = 330 \text{ }\Omega$; $R_L = 1.5 \text{ k}\Omega$ (AC load between output pins)	–	15	22	$\frac{\text{nV}}{\sqrt{\text{Hz}}}$
F	noise figure of AM mixer 2		–	16	19.5	dB
AM IF2 AGC STAGE: PINS AMIF2IN AND AMIF2DEC						
$G_{\text{AGC}(\text{max})}$	maximum AGC amplifier gain	no AGC active	38	41	44	dB
$V_{\text{AGC}(\text{start})}$	AGC start voltage (pin IFOUT)	–3 dB below maximum output level; $R_L = 1 \text{ k}\Omega$	93	–	98	$\text{dB}\mu\text{V}$
ΔAGC	AGC range	between start and stop of AGC	52	–	–	dB
THD	total harmonic distortion of IF2 output signal	AM signal; $f_{\text{mod}} = 400 \text{ Hz}$; $m = 0.8$; $C_{\text{TAMIF2AGC}} = 4.7 \text{ }\mu\text{F}$; $V_{\text{AMIF2IN-AMIF2DEC}} < 0.14 \text{ V}$	–	–	0.3	%
$V_{i(n)(\text{eq})}$	equivalent input noise voltage	$R_{\text{gen}} = 1.5 \text{ k}\Omega$	–	10.5	–	$\frac{\text{nV}}{\sqrt{\text{Hz}}}$
R_i	input resistance		1.3	1.5	1.65	$\text{k}\Omega$
C_i	input capacitance		–	–	5	pF
AM IF2 AGC STAGE: PIN IFOUT						
$V_{o(\text{rms})}$	AM IF2 output voltage (RMS value)	$R_L = 1 \text{ k}\Omega$; AGC active	63	84	112	mV
t_{st}	AM IF2 AGC settling time	$V_{\text{AMIF2IN}} = 10 \text{ to } 100 \text{ mV}$	–	60	–	ms
		$V_{\text{AMIF2IN}} = 100 \text{ to } 10 \text{ mV}$	–	360	–	ms
		$V_{\text{AMIF2IN}} = 140 \text{ }\mu\text{V to } 140 \text{ mV}$	–	150	–	ms
		$V_{\text{AMIF2IN}} = 140 \text{ mV to } 140 \text{ }\mu\text{V}$	–	600	–	ms

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
AM IF2 LEVEL DETECTOR OUTPUT: PIN VLEVEL; see Fig.4; note 5						
V _{VLEVEL}	DC output voltage	V _i = 10 μV to 1 V	0	–	5	V
		V _i < 1 μV; standard setting of level DAA	0.1	0.35	0.8	V
		V _i = 1.4 mV; standard setting of level DAA	1.6	2.2	2.8	V
ΔV _{level}	step size for adjustment of level starting point	V _i = 0 V; standard setting of level slope	40	53	70	mV
V _{level(slope)}	slope of level voltage $\frac{\Delta V_{\text{level}}}{\Delta V_{\text{AMIF2IN-AMIF2DEC}}}$	V _i = 140 μV to 140 mV; standard setting of level slope	0.8	1	1.2	$\frac{\text{V}}{20 \text{ dB}}$
ΔV _{step}	step size for adjustment of level slope	V _i = 1.4 mV	55	75	95	$\frac{\text{mV}}{20 \text{ dB}}$
B _{level}	bandwidth of level output voltage	V _i = 15 mV; standard setting of level DAA	200	300	–	kHz
R _o	output resistance		–	–	500	Ω
RR	ripple rejection $\frac{V_{\text{VLEVEL}}}{V_{\text{DDA6}}}$	V _{DDA6(ripple)} = 100 mV; f _{ripple} = 100 Hz	–	40	–	dB
AM NOISE BLANKER; AMNB = 1; TEST SIGNAL AND TEST CIRCUIT; see Fig.3						
t _{sup}	suppression time	V _{pulse} = 250 mV; V _{VLEVEL} < 1.8 V	6	7.5	10	μs
f _{trigger}	trigger sensitivity frequency	V _{pulse} = 200 mV; V _{VLEVEL} < 1.8 V	–	1000	–	Hz
		V _{pulse} = 200 mV; V _{VLEVEL} > 2.2 V	–	–	100	Hz
		V _{pulse} = 20 mV; V _{VLEVEL} < 1.8 V	–	–	100	Hz
I _{sink(AGC)}	AM noise blanker AGC sink current	V _{TAMNBAGC} = 3 V	35	50	65	μA
V _{AGC}	AM noise blanker AGC voltage (pin TAMNBAGC)	AM mixer 1 input V _i = 0 V	2	2.4	2.8	V
FM signal channel						
FM RF AGC (FM DISTANCE MODE; LODX = 0)						
RF input: pins FMMIXIN1 and FMMIXIN2; KAGC = 0						
V _{i(RF)}	RF input voltage for start of wideband AGC	AGC[1:0] = 11	–	3	–	mV
		AGC[1:0] = 10	–	6	–	mV
		AGC[1:0] = 01	–	9	–	mV
		AGC[1:0] = 00	–	12	–	mV

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
<i>FM RF AGC time constant: pin TFMAGC</i>						
R_{source}	source resistance		4	5	6	$k\Omega$
$V_{O(ref)}$	DC output reference voltage	AGC[1:0] = 00; KAGC = 0; $V_{FMMIXIN1-FMMIXIN2} = 0\text{ V}$	3.6	4.1	4.6	V
<i>FM RF AGC PIN diode drive output: pin IFMAGC</i>						
$I_{sink(max)}$	maximum AGC sink current	$V_{IFMAGC} = 2.5\text{ V};$ $V_{TFMAGC} = V_{O(ref)} - 0.5\text{ V};$ AGC[1:0] = 00; KAGC = 0	8	11.5	15	mA
$I_{source(max)}$	maximum AGC source current	$V_{IFMAGC} = 2.5\text{ V};$ $V_{TFMAGC} = V_{O(ref)} + 0.5\text{ V};$ AGC[1:0] = 00; KAGC = 0	-15	-11.5	-8	mA
$I_{source(AGC)}$	AGC source current	AM mode	-15	-11.5	-8	mA
		$V_{IFMAGC} = 2.5\text{ V}; LODX = 1$	-4.4	-3.7	-2.7	mA
<i>FM keyed AGC: pin VLEVEL</i>						
V_{th}	threshold voltage for narrow-band AGC	KAGC = 1; $V_{TFMAGC} = V_{O(ref)} + 0.6\text{ V}$	0.6	0.75	0.9	V
<i>FM RF AGC TRANSCONDUCTANCE BUFFER: PIN AGCBUF</i>						
$g_{m(buf)}$	buffer transconductance	FM mode; $V_{TFMAGC} = V_{O(ref)}$ to $V_{O(ref)} + 80\text{ mV}$	4	5.3	6.4	mS
$I_{sink(max)}$	maximum sink current	FM mode; open-collector; $V_{TFMAGC} = V_{O(ref)} + 0.15\text{ V}$	450	500	560	μA
$I_{source(max)}$	maximum source current	FM mode; $V_{TFMAGC} = V_{O(ref)}$	-	-	-30	μA
$V_{o(n)}$	buffer output noise voltage	FM mode; $V_{VDDA2-AGCBUF} = 1\text{ V}$ (voltage across external pull-up resistor); B = 400 Hz to 20 kHz	-	10	15	μV
<i>FM RF MIXER</i>						
<i>Reference voltage: pin VREFFMMIX</i>						
V_{ref}	reference voltage	FM mode	6.6	7.3	8	V
		AM mode	2.7	3.1	3.4	V
<i>Inputs: pins FMMIXIN1 and FMMIXIN2</i>						
R_i	input resistance	MIXGAIN = 0	-	3.5	-	$k\Omega$
		MIXGAIN = 1	-	1.8	-	$k\Omega$
C_i	input capacitance		-	5	7	pF
V_{BIAS}	DC bias voltage	FM mode	2.2	2.7	3.2	V
$V_{i(RF)(p)}$	RF input voltage (peak value)	1 dB compression point of FM mixer output				
		MIXGAIN = 0	70	100	-	mV
		MIXGAIN = 1	35	50	-	mV

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<i>Outputs: pins MIX1OUT1 and MIX1OUT2</i>						
R _o	output resistance		100	–	–	kΩ
C _o	output capacitance		2	3.5	5	pF
V _{o(max)(p-p)}	maximum output voltage (peak-to-peak value)		3	–	–	V
I _{bias}	mixer bias current	FM mode	4.8	6	7.2	mA
<i>FM mixer</i>						
g _{m(conv)}	conversion transconductance $\frac{I_{IF}}{V_{RF}}$	MIXGAIN = 0	8.5	12.5	18	$\frac{mA}{V}$
		MIXGAIN = 1	17	25	36	$\frac{mA}{V}$
g _{m(conv)(T)}	conversion transconductance variation with temperature $\frac{\Delta g_{m(conv)}}{g_{m(conv)} \times \Delta T}$	MIXGAIN = 0	–	-1×10^{-3}	–	K ⁻¹
F	noise figure	MIXGAIN = 0	–	3.5	4.6	dB
		MIXGAIN = 1	–	2.4	–	dB
IP3	3rd-order intermodulation	MIXGAIN = 0	113	117	–	dBμV
		MIXGAIN = 1	–	108	–	dBμV
IRR	image rejection ratio	FMINJ = 1; note 6 f _{RFwanted} = 87.5 MHz; f _{RFimage} = 108.9 MHz data byte 3 = X010X110;	25	30	–	dB
		f _{RFwanted} = 162.475 MHz; f _{RFimage} = 183.875 MHz	22	30	–	dB
V _{i(n)(eq)}	equivalent input noise voltage (pin FMMIXIN1 to FMMIXIN2)	R _{gen} = 200 Ω; R _L = 2.6 kΩ MIXGAIN = 0	–	2.9	3.1	$\frac{nV}{\sqrt{Hz}}$
		MIXGAIN = 1	–	2.6	–	$\frac{nV}{\sqrt{Hz}}$
R _{gen(opt)}	optimum generator resistance		–	200	–	Ω

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
FM IF AMPLIFIER (IBOC = 0)						
G	gain	$R_L = 330 \Omega$; $V_i = 1 \text{ mV}$; note 7				
		IFGAIN = 0	–	10.5	–	dB
F	noise figure	IFGAIN = 0	–	10	–	dB
		IFGAIN = 1	–	8.3	–	dB
IP3	3rd-order intermodulation	IFGAIN = 0	–	123	–	dB μ V
		IFGAIN = 1	–	119	–	dB μ V
<i>Inputs: pins IFAMPIN and IFAMPDEC</i>						
$V_{i(\text{max})(p)}$	maximum input voltage (peak value)	1 dB compression point of FM IF amplifier output voltage				
		IFGAIN = 0	400	–	–	mV
$V_{i(n)(\text{eq})}$	equivalent input noise voltage	$R_{\text{gen}} = 330 \Omega$				
		IFGAIN = 0	–	8	10	$\frac{\text{nV}}{\sqrt{\text{Hz}}}$
$V_{i(n)(\text{eq})}$	equivalent input noise voltage	IFGAIN = 1	–	6.5	10	$\frac{\text{nV}}{\sqrt{\text{Hz}}}$
$V_{i(\text{IFAMP})}$	DC voltage IF amplifier input	FM mode	2.2	2.7	3.2	V
R_i	input resistance		270	330	390	Ω
C_i	input capacitance		–	5	7	pF
<i>Output: pin FMIFAMPOUT</i>						
$V_{o(\text{max})(p)}$	maximum output voltage (peak value)		1.2	1.5	–	V
$V_{O(\text{IFAMP})}$	DC voltage IF amplifier output	FM mode	3	3.5	4	V
R_o	output resistance		270	330	390	Ω
C_o	output capacitance		–	5	7	pF
IBOC AMPLIFIER (IBOC = 1)						
G	gain		–	8	–	dB
IP3	3rd-order intermodulation		–	123	–	dB μ V
<i>Inputs: pins IFAMPIN and IFAMPDEC</i>						
$V_{i(n)(\text{eq})}$	equivalent input noise voltage	$R_{\text{gen}} = 330 \Omega$	–	5.2	10	$\frac{\text{nV}}{\sqrt{\text{Hz}}}$
$V_{i(\text{max})(p)}$	maximum input voltage (peak value)	1 dB compression point of FM IF amplifier output voltage	420	–	–	mV

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
FM IF AGC						
G_{AGC}	small signal AGC amplifier gain $\frac{V_{IFOUT}}{V_{FMIFIN-FMIFDEC}}$	linear range: $V_{FMIFIN-FMIFDEC} = 25 \mu\text{V}$ to 1 mV; no AGC active	–	28.5	–	dB
ΔAGC	AGC range		59	–	–	dB
IP3	3rd-order intermodulation	$V_{FMIFIN-FMIFDEC} = 95 \text{ dB}\mu\text{V}$	–	121	–	dB μV
t_{st}	FM IF AGC settling time	$V_{FMIFIN} = 1$ to 10 mV	–	2	–	ms
		$V_{FMIFIN} = 10$ to 1 mV	–	1	–	ms
<i>Inputs: pins FMIFIN and FMIFDEC</i>						
$V_{i(IF)}$	IF input voltage for IF AGC start level	start of AGC	–	65	–	dB μV
$V_{i(max)(p)}$	maximum input voltage (peak value)		–	1.6	–	V
$V_{i(n)(eq)}$	equivalent input noise voltage	$R_{gen} = 330 \Omega$; no AGC active	–	7	–	$\frac{\text{nV}}{\sqrt{\text{Hz}}}$
R_i	input resistance		270	330	390	Ω
C_i	input capacitance		–	5	7	pF
<i>Output: pin IFOUT</i>						
V_O	FM IF output voltage	$R_L = 1 \text{ k}\Omega$; AGC active	89	94	99	dB μV
FM IF LEVEL DETECTOR OUTPUT: PIN VLEVEL; see Fig.5; note 8						
V_{VLEVEL}	DC output voltage	$V_i = 10 \mu\text{V}$ to 1 V	0	–	5	V
		$V_i < 1 \mu\text{V}$; standard setting of level DAA	0.1	0.35	0.8	V
		$V_i = 1 \text{ mV}$; standard setting of level DAA	1.4	2	2.4	V
ΔV_{level}	step size for adjustment of level starting point	LSL[2:0] = 100	40	53	70	mV
$V_{level(slope)}$	slope of level voltage $\frac{\Delta V_{level}}{\Delta V_i}$	$V_i = 1$ to 300 mV; standard setting of level slope	0.8	1	1.2	$\frac{\text{V}}{20 \text{ dB}}$
ΔV_{step}	step size for adjustment of level slope	$V_i = 1 \text{ mV}$	55	75	95	$\frac{\text{mV}}{20 \text{ dB}}$
B_{level}	bandwidth of level output voltage	$V_i = 10 \text{ mV}$; standard setting of level DAA	200	300	–	kHz
I_{source}	output source current	$V_{level(ref)} - 60 \text{ mV}$; $V_i = 1 \text{ mV}$	–	–	–300	μA
I_{sink}	output sink current	$V_{level(ref)} + 25 \text{ mV}$; $V_i = 1 \text{ mV}$	50	–	–	μA
R_o	output resistance		–	–	500	Ω
RR	ripple rejection $\frac{V_{VLEVEL}}{V_{DDA6}}$	$V_{DDA6(ripple)} = 100 \text{ mV}$; $f_{ripple} = 100 \text{ Hz}$	–	40	–	dB

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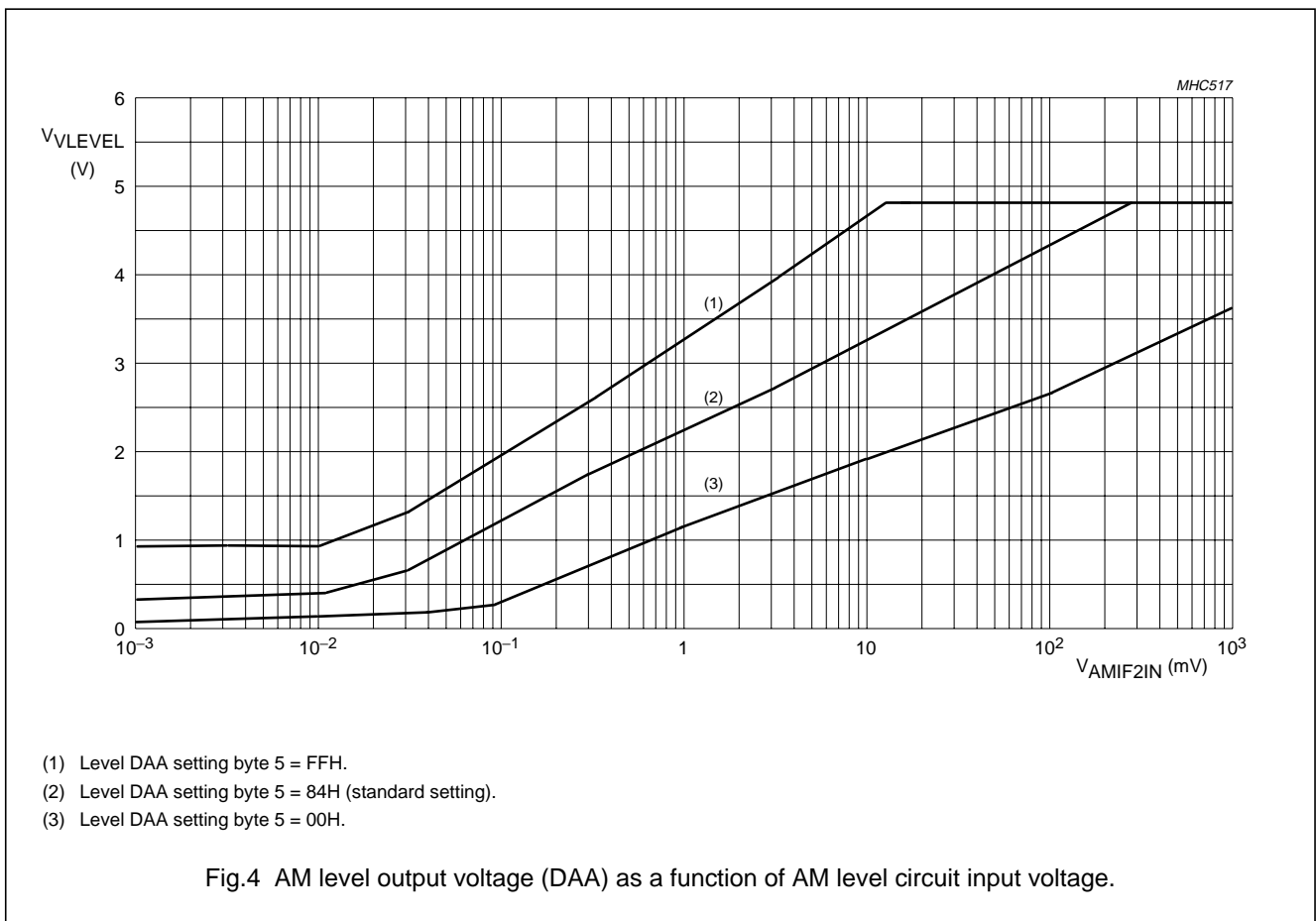
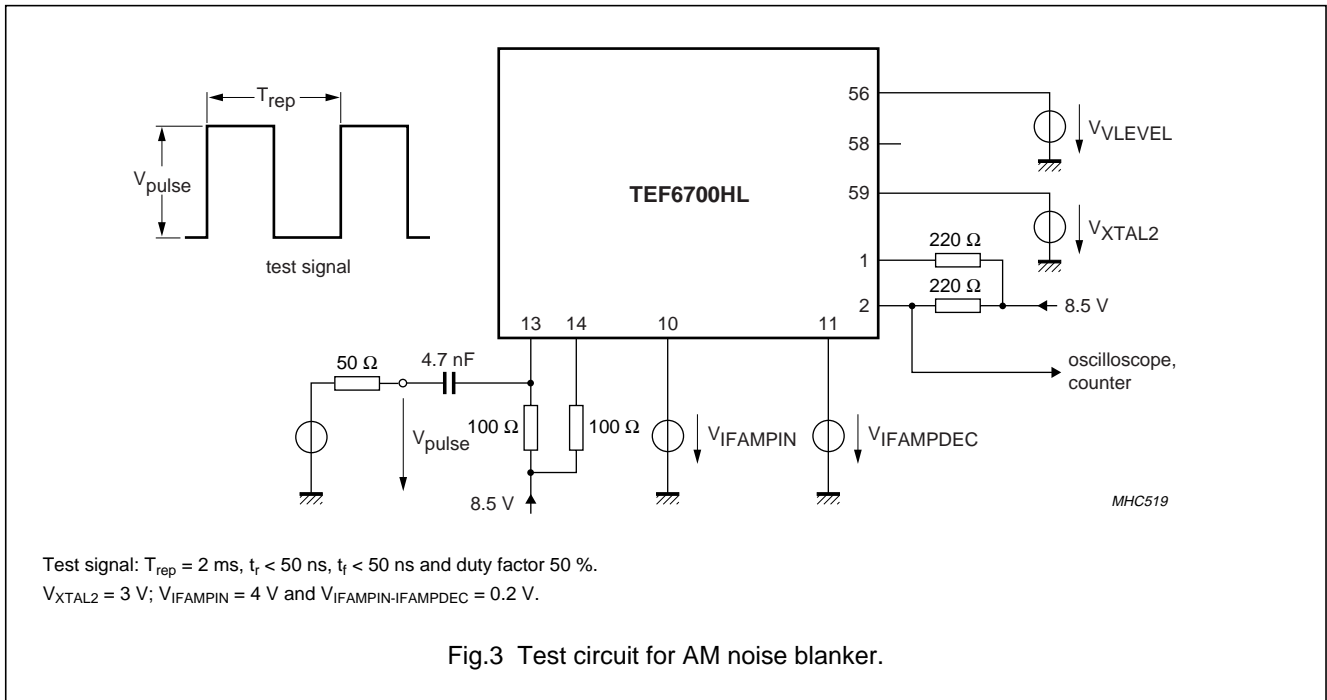
SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
FM IF/AM IF2 buffer						
<i>Output: pin IFOUT</i>						
$Z_{L(\min)}$	minimum load impedance for IF multiplexer/buffer		1	–	–	k Ω
$V_{o(\max)(p)}$	maximum output voltage (peak value)		–	–	750	mV
RDS update						
<i>Output: pin AFHOLD; see Fig.6</i>						
$I_{\text{sink}(\max)}$	maximum sink current	after first bus transmission with AF = 1; $V_o = 0.5$ V	1	1.2	1.4	mA

Notes

- Measured between pins XTAL1 and XTAL2.
- DAA conversion gain formula: $V_{\text{DAAOUT}} = \left[2 \times \left(0.75 \times \frac{n}{128} + 0.125 \right) \times (V_{\text{DAAIN}} + V_{\text{DAATD}}) \right] - V_{\text{DAATD}}$; where $n = 0$ to 127.
- $V_{\text{LSB}} = V_{\text{DAAOUT}(n+1)} - V_{\text{DAAOUT}(n)}$
- The AM AGC transconductance buffer delivers a sink current which is proportional to the voltage change at pin TRFAMAGC. $\Delta V_{\text{TRFAMAGC}} = V_{\text{TRFAMAGC}} - V_{\text{TRFAMAGC}}|_{(V_{\text{AMMIX1IN}} - V_{\text{AMMIX1DEC}}) < 10 \text{ mV}}$
- Input parameters of AM IF2 measured between pins AMIF2IN and AMIF2DEC.
- Image rejection ratio: $\text{IRR} = \frac{V_{(\text{MIX1OUT1} - \text{MIX1OUT2})\text{wanted}}}{V_{(\text{MIX1OUT1} - \text{MIX1OUT2})\text{image}}}$
- Input parameters of FM IF amplifier measured between pins IFAMPIN and IFAMPDEC.
- Input parameters of FM IF level detector measured between pins FMIFIN and FMIFDEC.

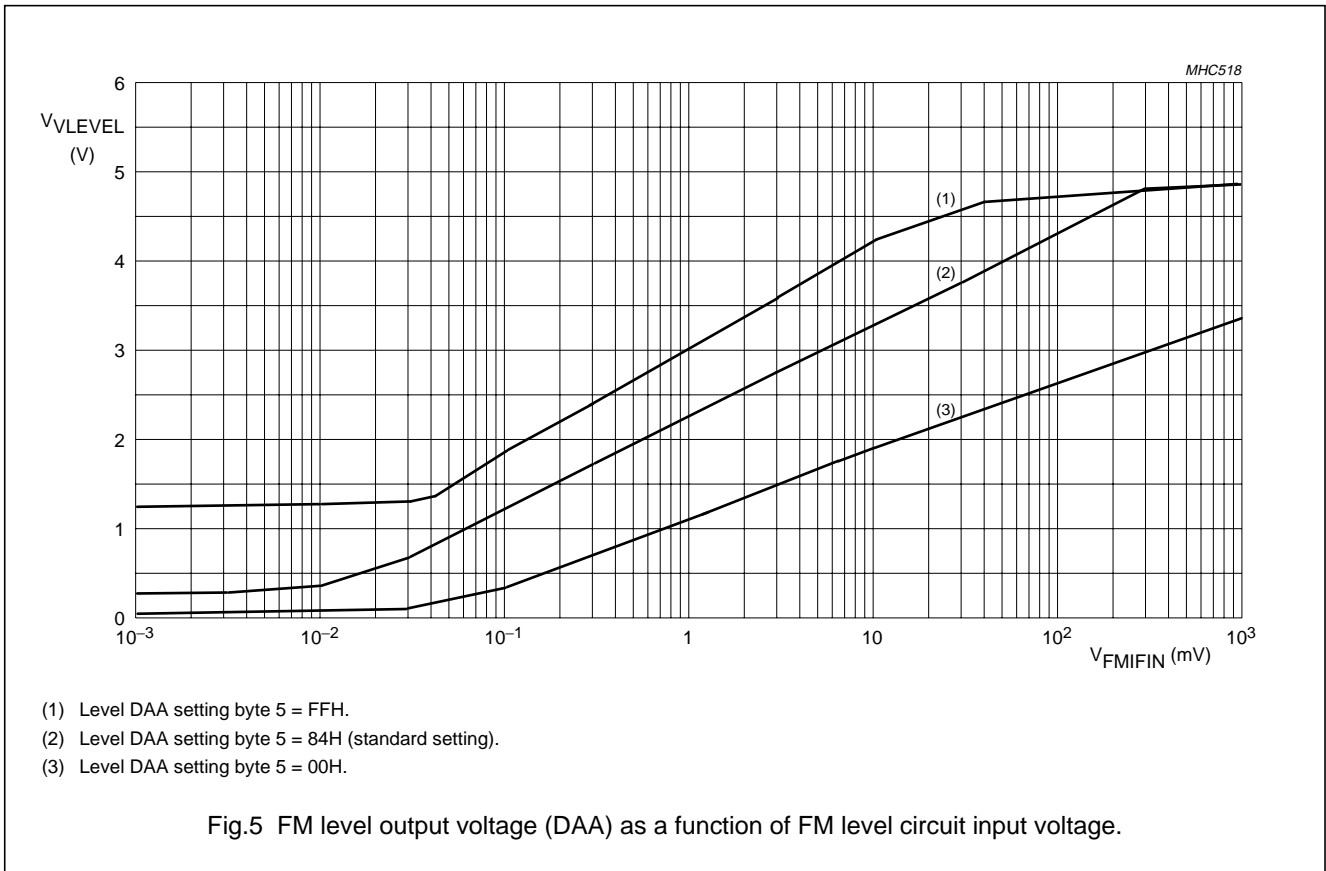
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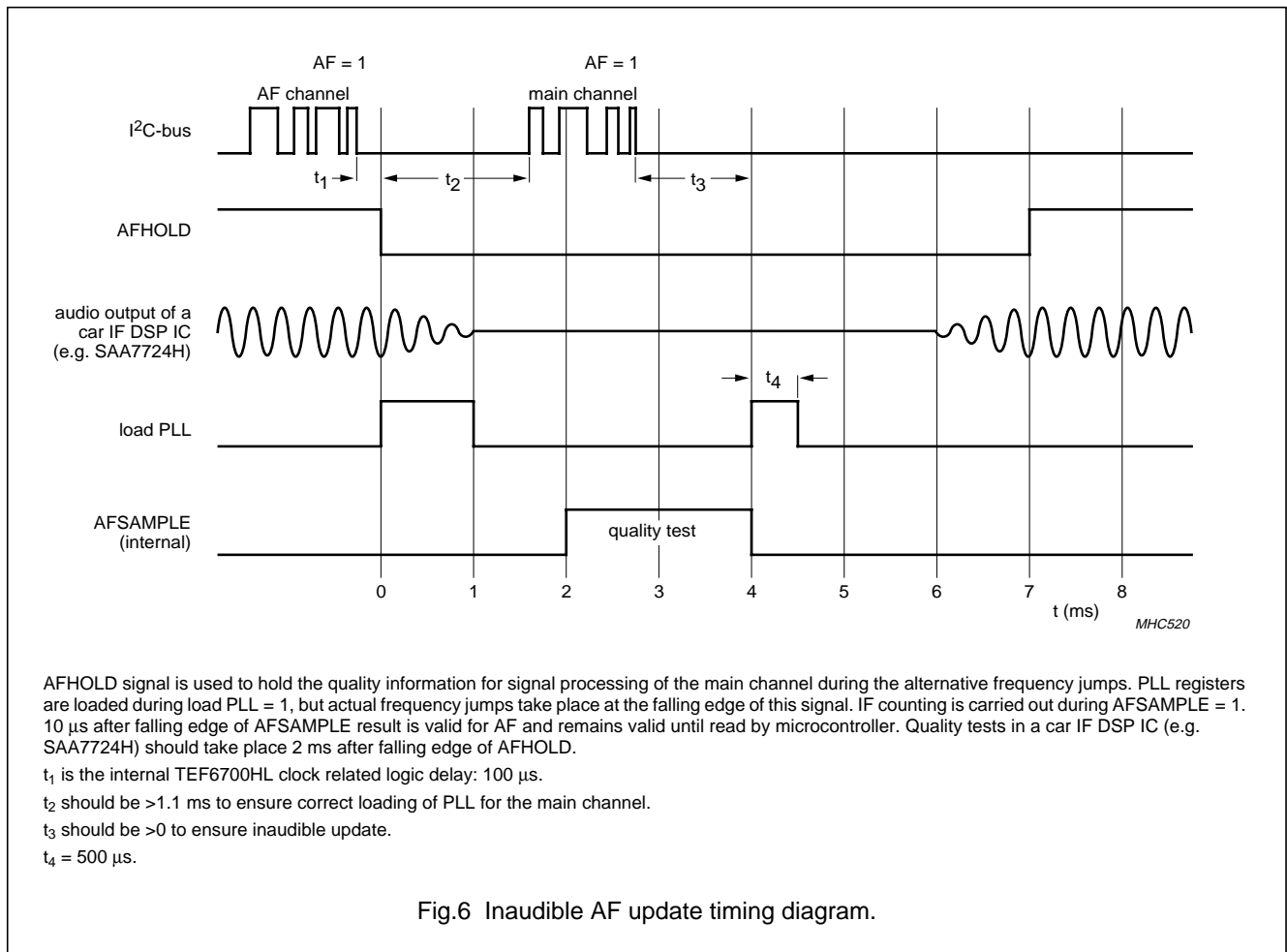
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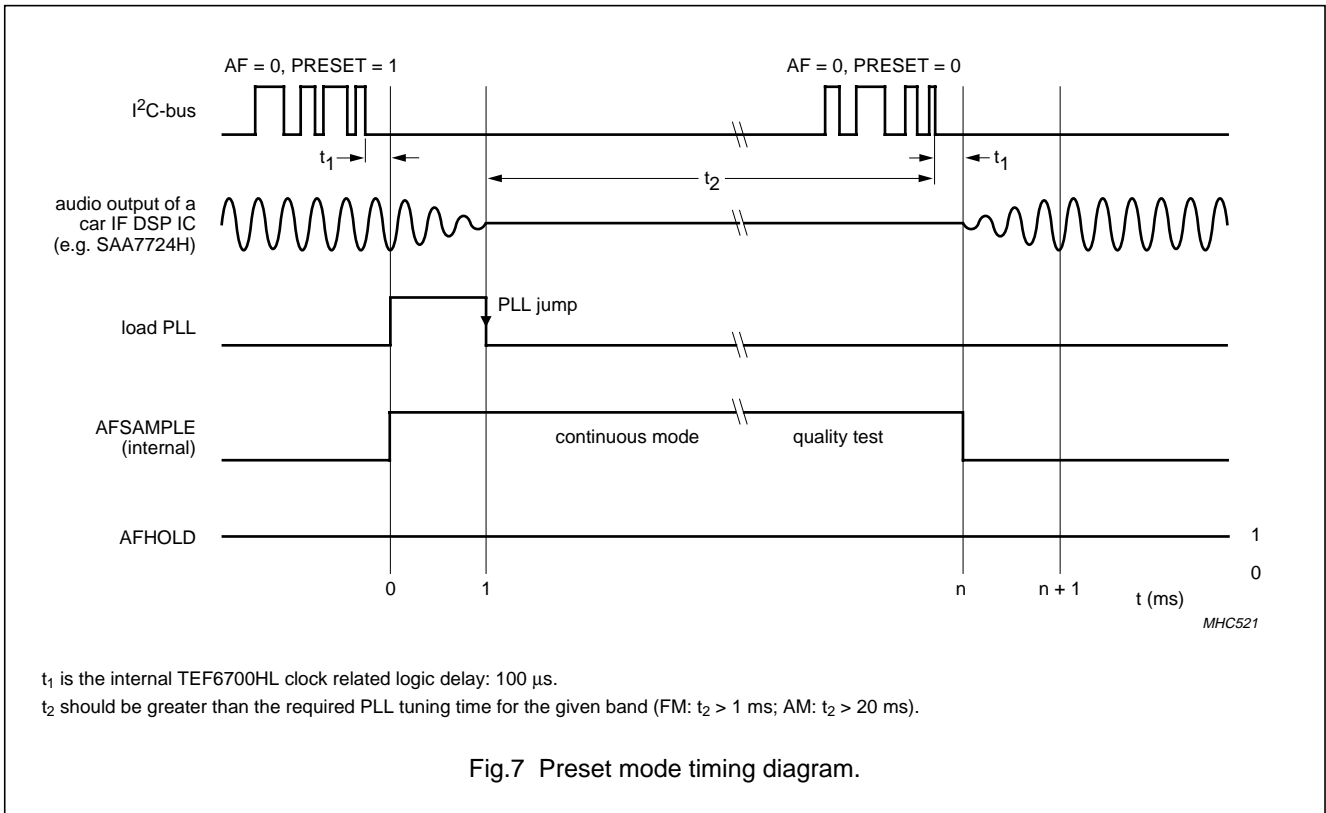
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12 I²C-BUS PROTOCOL**12.1 I²C-bus specification**

SDA and SCL HIGH and LOW levels are specified according to a 3.3 V I²C-bus. The bus pins tolerate also thresholds of a 5 V bus.

The standard I²C-bus specification is expanded by the following definitions.

IC addresses:

- 1st IC address C2H: 1100001 R/ \bar{W}
- 2nd IC address C0H: 1100000 R/ \bar{W}
- 3rd IC address C4H: 1100010 R/ \bar{W} .

Structure of the I²C-bus logic: slave transceiver with auto increment.

Subaddresses are not used.

The second I²C-bus address can be selected by connecting pin ADDRSEL via a 120 k Ω resistor to GND. The third I²C-bus address can be selected by connecting pin ADDRSEL via a 33 k Ω resistor to GND.

The maximum bit rate for this device is 100 kbits/s.

The I²C-bus interface is extended with an enable input (pin BUSENABLE). If pin BUSENABLE is HIGH the communication with the device is active; if pin BUSENABLE is LOW the signals on the I²C-bus are ignored so that higher bit rates (>100 kbits/s) can be used to communicate with other devices on the same I²C-bus. The enable signal must not change while bus communication takes place.

The data hold time $t_{HD,DAT}$ must be at least 1 μ s.

No default settings at power-on reset. I²C-bus transmission is required to program the IC.

12.1.1 DATA TRANSFER

Data sequence: address, byte 0, byte 1, byte 2, byte 3, byte 4, byte 5, and byte 6.

The data transfer has to be in this order. The LSB of the address being logic 0 indicates a write operation.

Bit 7 of each byte is considered the MSB and has to be transferred as the first bit of the byte.

The data becomes valid at the output of the internal latches with the acknowledge of each byte. A STOP condition after any byte can shorten transmission times.

When writing to the transceiver by using the STOP condition before completion of the whole transfer:

- The remaining bytes will contain the old information
- If the transfer of a byte is not completed, this byte is lost and the previous information is available.

12.1.2 FREQUENCY SETTING

For new frequency setting, in both AM and FM mode, the programmable divider is enabled by setting bit PRESET to logic 1. To select a frequency, two I²C-bus transmissions are necessary:

- First: bit PRESET = 1
- Second: bit PRESET = 0.

12.2 I²C-bus protocol**12.2.1 DATA TRANSFER MODE AND IC ADDRESS****Table 1** Write mode

S ⁽¹⁾	address (write)	A ⁽²⁾	data byte(s)	A ⁽²⁾	P ⁽³⁾
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Table 2 Read mode

S ⁽¹⁾	address (read)	A ⁽²⁾	data byte 0	NA ⁽⁴⁾	P ⁽³⁾
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Notes

1. S = START condition.
2. A = acknowledge.
3. P = STOP condition.
4. NA = no acknowledge.

Table 3 IC address byte

ADDRESS	IC ADDRESS ⁽¹⁾							MODE ⁽²⁾
1	1	1	0	0	0	0	1	R/ \bar{W}
2	1	1	0	0	0	0	0	R/ \bar{W}
3	1	1	0	0	0	1	0	R/ \bar{W}

Notes

1. Pin ADDRSEL left open-circuit activates 1st IC address; R_{ext} = 120 k Ω at pin ADDRSEL to ground activates 2nd IC address; R_{ext} = 33 k Ω at pin ADDRSEL to ground activates 3rd IC address.
2. Read or write bit:
0 = write operation to TEF6700HL
1 = read operation from TEF6700HL.

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12.2.2 WRITE MODE: DATA BYTE 0

Table 4 Format of data byte 0

BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
AF	PLL14	PLL13	PLL12	PLL11	PLL10	PLL9	PLL8

Table 5 Description of data byte 0 bits

BIT	SYMBOL	DESCRIPTION
7	AF	Alternative frequency. If AF = 0, then normal operation. If AF = 1, then AF (RDS) update mode.
6 to 0	PLL[14:8]	Setting of programmable counter of synthesizer PLL. Upper byte of PLL divider word.

12.2.3 WRITE MODE: DATA BYTE 1

Table 6 Format of data byte 1

BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
PLL7	PLL6	PLL5	PLL4	PLL3	PLL2	PLL1	PLL0

Table 7 Description of data byte 1 bits

BIT	SYMBOL	DESCRIPTION
7 to 0	PLL[7:0]	Setting of programmable counter of synthesizer PLL. Lower byte of PLL divider word.

12.2.4 WRITE MODE: DATA BYTE 2

Table 8 Format of data byte 2

BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
PRESET	DAA6	DAA5	DAA4	DAA3	DAA2	DAA1	DAA0

Table 9 Description of data byte 2 bits

BIT	SYMBOL	DESCRIPTION
7	PRESET	Preset. If PRESET = 0, then programmable divider and antenna DAA locked. If PRESET = 1, then writing to programmable divider and antenna DAA enabled.
6 to 0	DAA[6:0]	Setting of antenna digital auto alignment.

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12.2.5 WRITE MODE: DATA BYTE 3

Table 10 Format of data byte 3

BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
IFMT	FREF2	FREF1	FREF0	IFPR	BND1	BND0	AMFM

Table 11 Description of data byte 3 bits

BIT	SYMBOL	DESCRIPTION
7	IFMT	IF measuring time. If IFMT = 0, then IF measuring time is 20 ms. If IFMT = 1, then IF measuring time is 2 ms.
6 to 4	FREF[2:0]	Reference frequency for synthesizer. These 3 bits determine the reference frequency, see Table 12.
3	IFPR	IF counter prescaler ratio. If IFPR = 0, then IF prescaler ratio is 100. If IFPR = 1, then IF prescaler ratio is 10.
2 and 1	BND[1:0]	Band switch. These 2 bits select the frequency in AM and FM mode, see Tables 13 and 14.
0	AMFM	AM or FM switch. If AMFM = 0, then FM mode. If AMFM = 1, then AM mode.

Table 12 Reference frequency setting

FREF2	FREF1	FREF0	f_{ref} (kHz)
0	0	0	100
1	0	0	50
0	1	0	25
1	1	0	20
0	0	1	10
1	0	1	10
0	1	1	10
1	1	1	10

Table 13 FM band selection bits

BND1	BND0	FREQUENCY BAND	VCO DIVIDER	CHARGE PUMP CURRENT
0	0	FM standard	2	130 μ A + 3 mA
0	1	FM Japan	3	130 μ A + 3 mA
1	0	FM East Europe	3	1 mA
1	1	FM weather	1	300 μ A

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Table 14 AM band selection bits; note 1

BND1	BND0	FREQUENCY BAND	VCO DIVIDER	CHARGE PUMP CURRENT (mA)
0	X	AM SW	10	1
1	X	AM LW/MW	20	1

Note

1. X = don't care.

12.2.6 WRITE MODE: DATA BYTE 4

Table 15 Format of data byte 4

BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
KAGC	AGC1	AGC0	LODX	FMINJ	IBOC	IFGAIN	MIXGAIN

Table 16 Description of data byte 4 bits

BIT	SYMBOL	DESCRIPTION
7	KAGC	Keyed FM AGC. If KAGC = 0, then keyed FM AGC is off. If KAGC = 1, then keyed FM AGC is on.
6 and 5	AGC[1:0]	Wideband AGC. These 2 bits set the start value of wideband AGC. For AM, see Table 17 and for FM, see Table 18.
4	LODX	Local or distance. If LODX = 0, then distance mode is on. If LODX = 1, then local mode is on.
3	FMINJ	FM mixer image rejection. If FMINJ = 0, then low injection. If FMINJ = 1, then high injection.
2	IBOC	FM IBOC mode. If IBOC = 0, then IBOC buffer amplifier is off. If IBOC = 1, then IBOC buffer amplifier is on.
1	IFGAIN	IF amplifier gain. If IFGAIN = 0, then $G_{IFAMP} = 10.5$ dB. If IFGAIN = 1, then $G_{IFAMP} = 14$ dB.
0	MIXGAIN	FM mixer gain. If MIXGAIN = 0, then the FM mixer gain is nominal. If MIXGAIN = 1, then the FM mixer gain is +6 dB.

Table 17 Setting of wideband AGC for AM ($m = 0.3$)

AGC1	AGC0	AM MIXER 1 INPUT VOLTAGE (PEAK VALUE) (mV)
0	0	520
0	1	680
1	0	840
1	1	980

Table 18 Setting of wideband AGC for FM

AGC1	AGC0	FM RF MIXER INPUT VOLTAGE (RMS VALUE) (mV)
1	1	3
1	0	6
0	1	9
0	0	12

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12.2.7 WRITE MODE: DATA BYTE 5

Table 19 Format of data byte 5

BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
LST4	LST3	LST2	LST1	LST0	LSL2	LSL1	LSL0

Table 20 Description of data byte 5 bits

BIT	SYMBOL	DESCRIPTION
7 to 3	LST[4:0]	Setting of level DAA starting point. These 5 bits determine the offset of the level detector output voltage.
2 to 0	LSL[2:0]	Setting of level DAA slope. These 3 bits determine the steepness of the level detector output voltage.

Table 21 Standard setting of data byte 5 bits

SETTING OF LEVEL DAA STARTING POINT					SETTING OF LEVEL DAA SLOPE		
LST4	LST3	LST2	LST1	LST0	LSL2	LSL1	LSL0
1	0	0	0	0	1	0	0

12.2.8 WRITE MODE: DATA BYTE 6

Table 22 Format of data byte 6

BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
SWPORT2	SWPORT1	AMNB	–	–	–	–	–

Table 23 Description of data byte 6 bits

BIT	SYMBOL	DESCRIPTION
7	SWPORT2	Software programmable port 2. If SWPORT2 = 0, then pin SWPORT2 is inactive (high-impedance). If SWPORT2 = 1, then pin SWPORT2 is active (pull down to ground).
6	SWPORT1	Software programmable port 1. If SWPORT1 = 0, then pin SWPORT1 is inactive (high-impedance). If SWPORT1 = 1, then pin SWPORT1 is active (pull down to ground).
5	AMNB	AM noise blanker. If AMNB = 0, then the AM noise blanker is off. If AMNB = 1, then the AM noise blanker is on.
4 to 0	–	These 5 bits are not used and should be set to logic 0.

12.2.9 READ MODE: DATA BYTE 0

Table 24 Format of 1st data byte

BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
IFC7	IFC6	IFC5	IFC4	IFC3	IFC2	IFC1	IFC0

Table 25 Description of data byte 0 bits

BIT	SYMBOL	DESCRIPTION
7 to 0	IFC[7:0]	IF counter result. These bits contain the least significant eight bits of the IF counter result.

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13 TEST AND APPLICATION INFORMATION

13.1 Test information

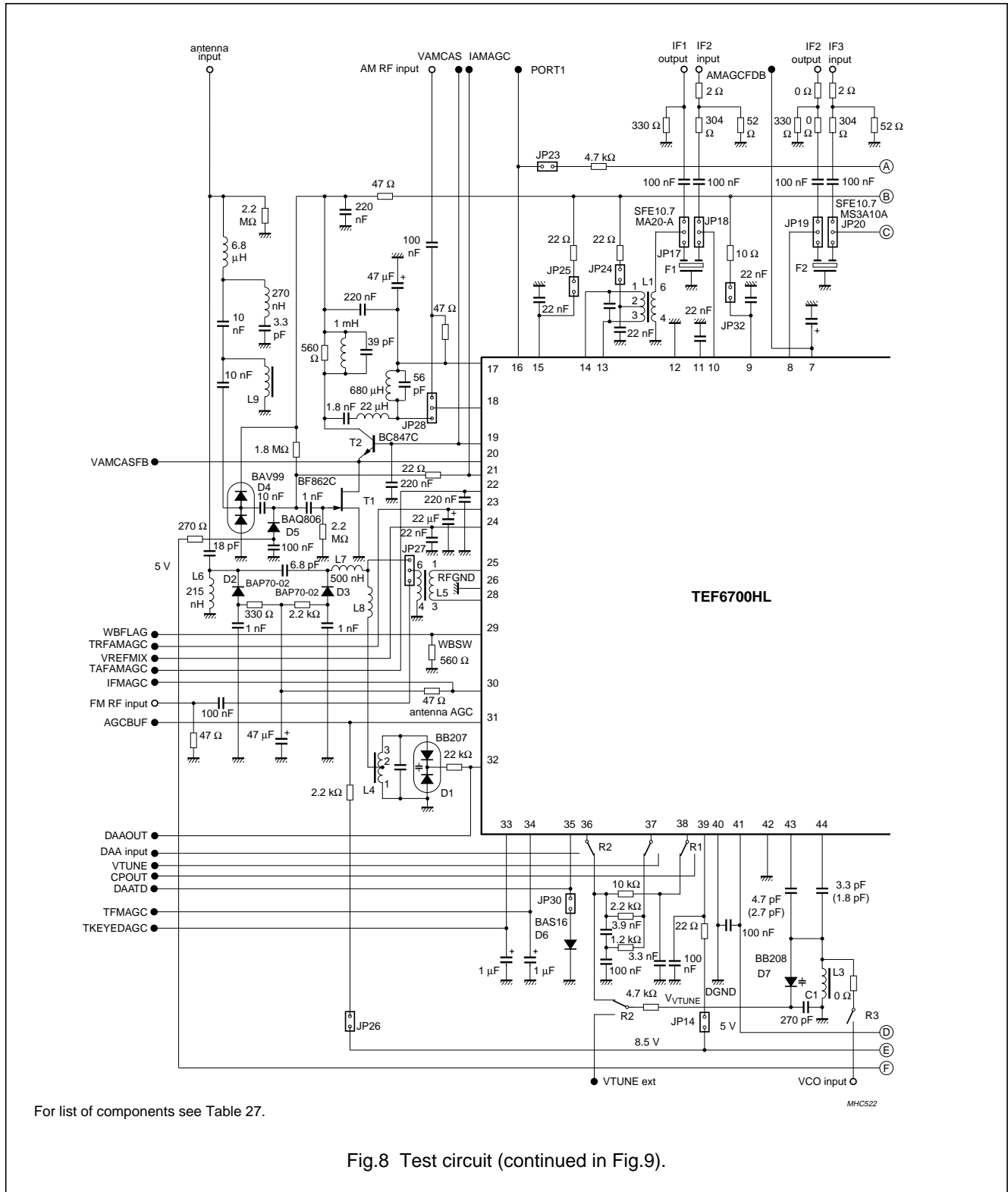


Fig.8 Test circuit (continued in Fig.9).

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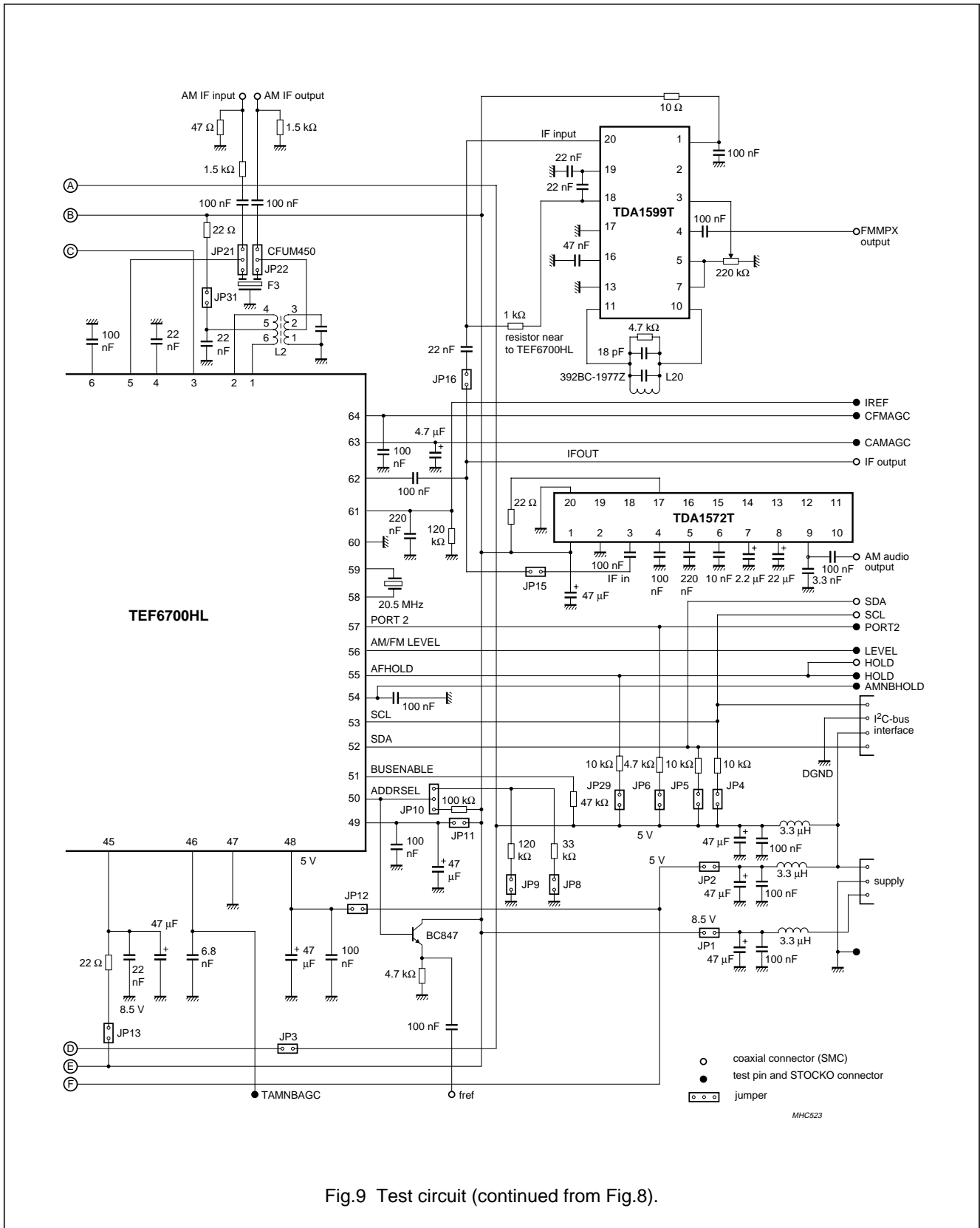


Fig.9 Test circuit (continued from Fig.8).

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Table 26 DC operating points

SYMBOL	PIN	UNLOADED DC VOLTAGE (V)					
		AM MODE			FM MODE		
		MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
AMMIX2OUT1	1	external 8.5			external 8.5		
AMMIX2OUT2	2	external 8.5			external 8.5		
FMIFIN	3	–	3.3	–	–	3.3	–
FMIFDEC	4	–	3.3	–	–	3.3	–
AMIF2IN	5	–	3	–	floating		
AMIF2DEC	6	–	3	–	floating		
i.c.	7	–	5.6	–	–	5.6	–
FMIFAMPOUT	8	–	8.1	–	–	4	–
V _{DDA1}	9	external 8.5			external 8.5		
IFAMPIN	10	–	2.9	–	–	2.7	–
IFAMPDEC	11	–	2.9	–	–	2.7	–
IFAMPGND	12	external 0			external 0		
MIX1OUT1	13	external 8.5			external 8.5		
MIX1OUT2	14	external 8.5			external 8.5		
V _{DDA2}	15	external 8.5			external 8.5		
SWPORT1	16	open-collector			open-collector		
AMMIX1DEC	17	–	2.8	–	floating		
AMMIX1IN	18	external biasing			floating		
VAMCAS	19	–	4.8	–	0	0.1	0.2
VAMCASFB	20	–	4.1	–	0	0.1	1
IAMAGC	21	7.1 (external biasing)			external biasing		
TAFAMAGC	22	–	0.3	–	0 (no WB)	0.3 (no WB)	0.5 (no WB)
TRFAMAGC	23	–	2.9	–	floating		
VREFFMMIX	24	–	3.2	–	–	7.3	–
FMMIXIN1	25	–	1.65	–	–	2.75	–
RFGND	26	external 0			external 0		
i.c.	27	external 0			external 0		
FMMIXIN2	28	–	1.65	–	–	2.75	–
WBFLAG	29	0			4 (WB)	4.5 (WB)	5 (WB)
					– (FM)	<0.5 (FM)	– (FM)
IFMAGC	30	5 (external biasing)			0.1 (external biasing)	–	4 (external biasing)

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SYMBOL	PIN	UNLOADED DC VOLTAGE (V)					
		AM MODE			FM MODE		
		MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
AGCBUF	31	8.5 (external biasing)			8.5 (external biasing)		
DAAOUT	32	–	0.2	0.3	0.2	–	8.25
TKEYEDAGC	33	floating			1	–	7
TFMAGC	34	7.8	8.3	8.8	–	4	–
DAATD	35	floating			0.2	–	1.5
DAAIN	36	0	–	8.5	0	–	8.5
VTUNE	37	0	–	8.5	0	–	8.5
CPOUT	38	0	–	8.5	0	–	8.5
V _{DDA3}	39	external 8.5			external 8.5		
DGND	40	external 0			external 0		
V _{DDD}	41	external 5			external 5		
VCOGND	42	external 0			external 0		
OSCFDB	43	2.2	2.8	3.4	2.2	2.8	3.4
OSCTNK	44	5	6.1	7.2	5	6.1	7.2
V _{DDA4}	45	external 8.5			external 8.5		
TAMNBAGC	46	–	2.4	–	floating		
IFGND	47	external 0			external 0		
V _{DDA5}	48	external 5			external 5		
V _{DDA6}	49	external 8.5			external 8.5		
ADDRSEL	50	–	3.6	–	–	3.6	–
BUSENABLE	51	external biasing			external biasing		
SDA	52	external biasing			external biasing		
SCL	53	external biasing			external biasing		
AMNBHOLD	54	4.3	4.6	5.1	8	8.4	–
AFHOLD	55	open-collector			open-collector		
VLEVEL	56	0	–	5	0	–	5
SWPORT2	57	open-collector			open-collector		
XTAL1	58	1.7	2.1	2.5	1.7	2.1	2.5
XTAL2	59	1.7	2.1	2.5	1.7	2.1	2.5
XTALGND	60	external 0			external 0		
IREF	61	4	4.25	4.5	4	4.25	4.5
IFOUT	62	–	2.5	–	–	4.6	–
TAMIF2AGC	63	–	4	–	–	4.2	–
TFMIFAGC	64	–	5	–	–	4.8	–

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13.2 Application information

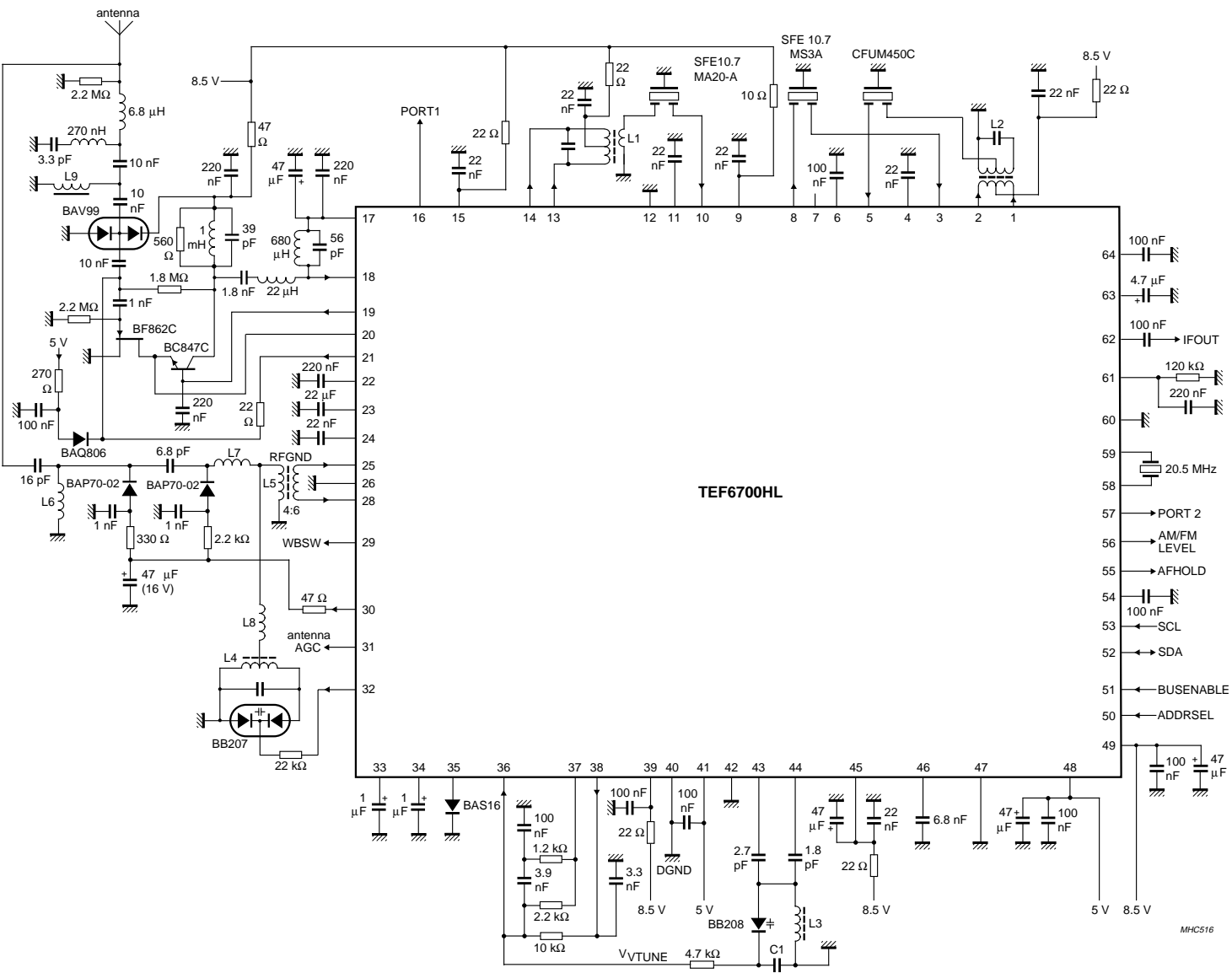


Fig.10 Application diagram.

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Table 27 List of components

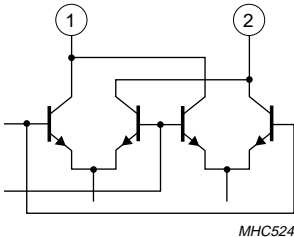
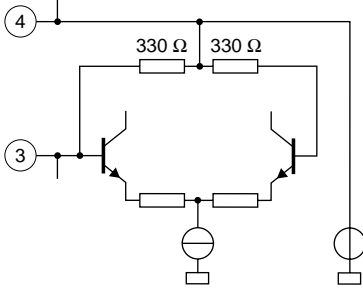
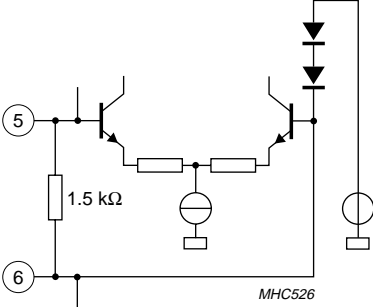
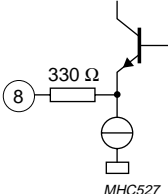
SYMBOL	PARAMETER	TYPE	MANUFACTURER
C1	capacitor for VCO tuning	270 pF; type NP0	–
L1	10.7 MHz IF coil	P7 PSG P826RC 5134N=S	TOKO
L2	450 kHz IF coil	P7PSGAE-5078D=S	TOKO
L3	oscillator coil	E543SNAS-02010	TOKO
L4	FM image rejection	611SNS-1066Y	TOKO
L5	FM input transformer	369INS-3076X	TOKO
L6	FM antenna coil	LQN1HR50; 215 nH	MURATA
L7	PIN diode bias	LQN1HR21; 500 nH	MURATA
L8	connection image reject	wire 10 mm/printed coil	–
L9	AM input	388BN-1211Z	TOKO
D1	double varicap diode	BB207	Philips
D2	silicon PIN diode	BAP70-02	Philips
D3	silicon PIN diode	BAP70-02	Philips
D4	high-speed double diode	BAV99	Philips
D5	AM PIN diode	BAQ806	Philips
D6	high-speed diode	BAS16	Philips
D7	varicap diode	BB208	Philips
T1	N-channel junction FET	BF862C	Philips
T2	NPN general purpose transistor	BC847C	Philips
F1	ceramic filter	10.7 MHz	MURATA
F2	ceramic filter	10.7 MHz	MURATA
F3	ceramic filter	450 kHz	MURATA
–	crystal 20.5 MHz	LN-G102-587	NDK

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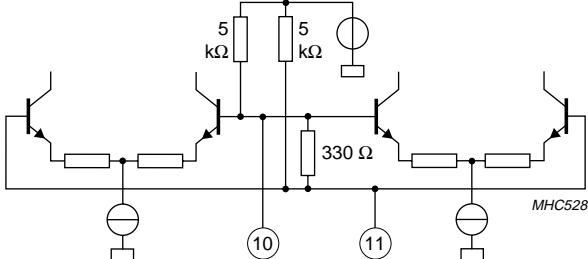
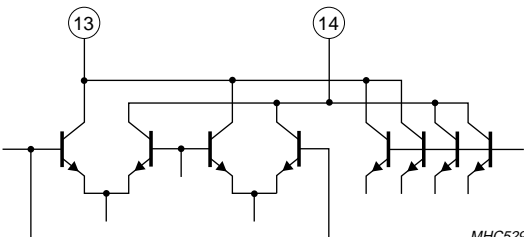
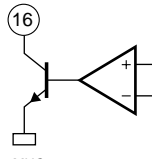
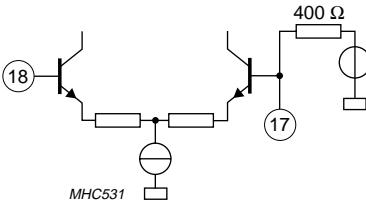
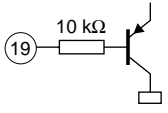
14 INTERNAL CIRCUITRY

Table 28 Equivalent pin circuits

PIN	SYMBOL	EQUIVALENT CIRCUIT
1	AMMIX2OUT1	 <p style="text-align: right; margin-right: 50px;">MHC524</p>
2	AMMIX2OUT2	
3	FMIFIN	 <p style="text-align: right; margin-right: 50px;">MHC525</p>
4	FMIFDEC	
5	AMIF2IN	 <p style="text-align: right; margin-right: 50px;">MHC526</p>
6	AMIF2DEC	
7	i.c.	
8	FMIFAMPOUT	 <p style="text-align: right; margin-right: 50px;">MHC527</p>
9	V _{DDA1}	

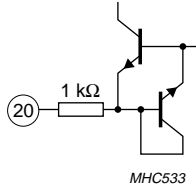
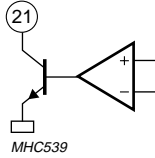
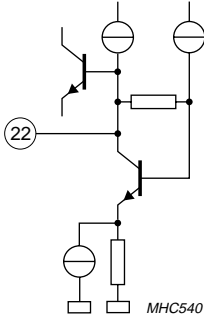
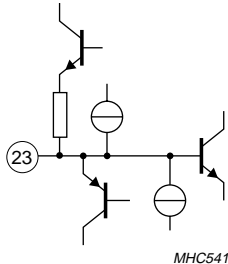
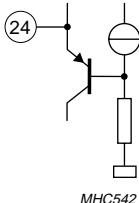
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PIN	SYMBOL	EQUIVALENT CIRCUIT
10	IFAMPIN	 <p style="text-align: right; margin-right: 20px;"><i>MHC528</i></p>
11	IFAMPDEC	
12	IFAMPGND	
13	MIX1OUT1	 <p style="text-align: right; margin-right: 20px;"><i>MHC529</i></p>
14	MIX1OUT2	
15	V _{DDA2}	
16	SWPORT1	 <p style="text-align: center;"><i>MHC530</i></p>
17	AMMIX1DEC	 <p style="text-align: center;"><i>MHC531</i></p>
18	AMMIX1IN	
19	VAMCAS	 <p style="text-align: center;"><i>MHC532</i></p>

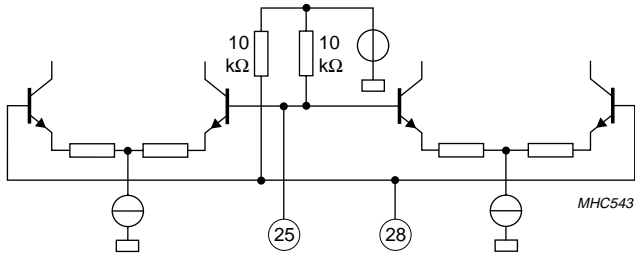
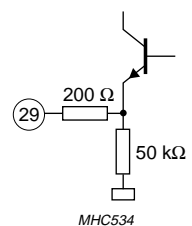
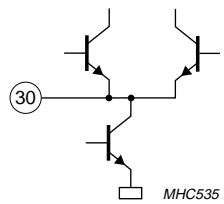
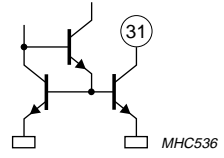
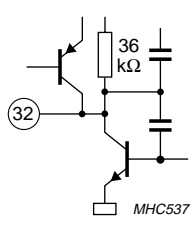
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PIN	SYMBOL	EQUIVALENT CIRCUIT
20	VAMCASFB	 <p>MHC533</p>
21	IAMAGC	 <p>MHC539</p>
22	TAFAMAGC	 <p>MHC540</p>
23	TRFAMAGC	 <p>MHC541</p>
24	VREFFMMIX	 <p>MHC542</p>

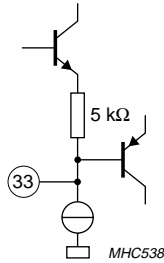
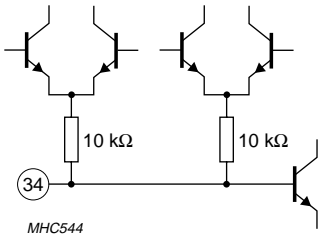
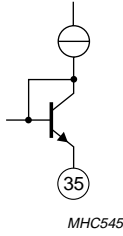
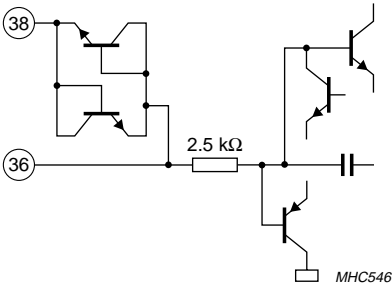
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PIN	SYMBOL	EQUIVALENT CIRCUIT
25	FMMIXIN1	
28	FMMIXIN2	
26	RFGND	
27	i.c.	
29	WBFLAG	
30	IFMAGC	
31	AGCBUF	
32	DAAOUT	

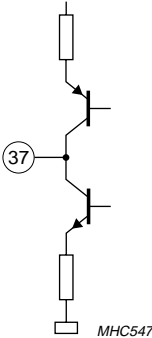
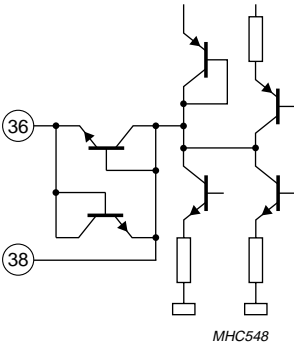
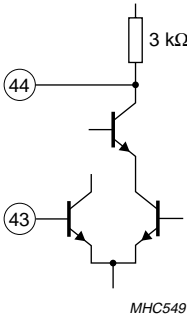
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PIN	SYMBOL	EQUIVALENT CIRCUIT
33	TKEYEDAGC	 <p>The diagram shows a transistor circuit. A 5 kΩ resistor is connected between the base and emitter of a transistor. The emitter is connected to a component labeled MHC538. The base is connected to pin 33. The collector is connected to another transistor.</p>
34	TFMAGC	 <p>The diagram shows two transistor stages. Each stage has a 10 kΩ resistor connected between the base and emitter. The emitters are connected to a component labeled MHC544. The bases are connected to pin 34. The collectors are connected to other transistors.</p>
35	DAATD	 <p>The diagram shows a transistor circuit. A component labeled MHC545 is connected between the base and emitter of a transistor. The emitter is connected to pin 35. The base is connected to another transistor.</p>
36	DAAIN	 <p>The diagram shows a complex transistor circuit. A 2.5 kΩ resistor is connected between the base and emitter of a transistor. The emitter is connected to a component labeled MHC546. The base is connected to pin 36. The collector is connected to another transistor. There are also other transistors and connections in the circuit.</p>

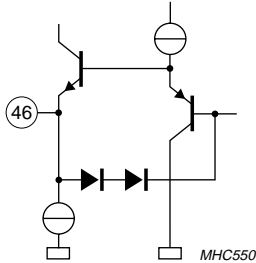
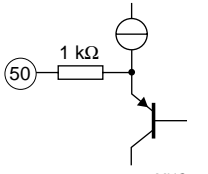
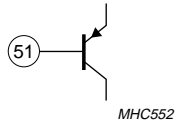
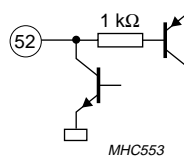
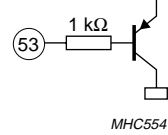
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PIN	SYMBOL	EQUIVALENT CIRCUIT
37	VTUNE	
38	CPOUT	
39	V_{DDA3}	
40	DGND	
41	V_{DDD}	
42	VCOGND	
43	OSCFDB	
44	OSCTNK	
45	V_{DDA4}	

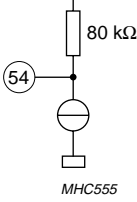
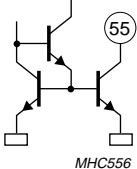
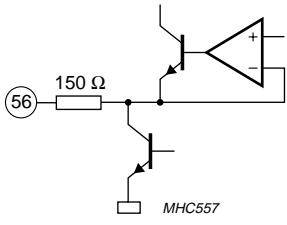
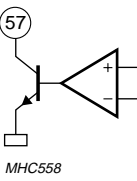
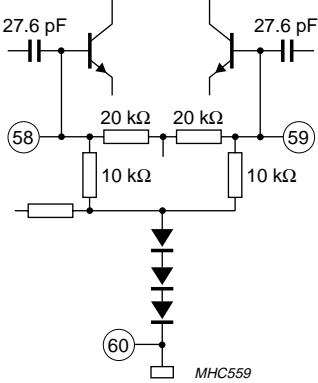
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PIN	SYMBOL	EQUIVALENT CIRCUIT
46	TAMNBAGC	 <p>MHC550</p>
47	IFGND	
48	V _{DDA5}	
49	V _{DDA6}	
50	ADDRSEL	 <p>MHC551</p>
51	BUSENABLE	 <p>MHC552</p>
52	SDA	 <p>MHC553</p>
53	SCL	 <p>MHC554</p>

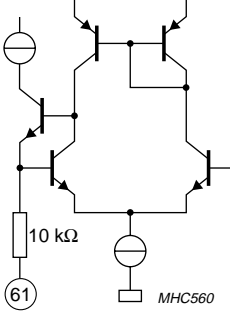
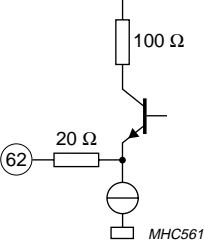
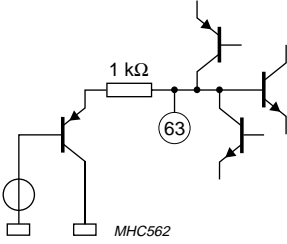
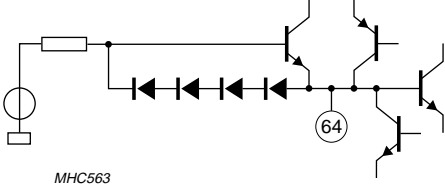
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PIN	SYMBOL	EQUIVALENT CIRCUIT
54	AMNBHOLD	
55	AFHOLD	
56	VLEVEL	
57	SWPORT2	
58	XTAL1	
59	XTAL2	
60	XTALGND	

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PIN	SYMBOL	EQUIVALENT CIRCUIT
61	IREF	 <p>The diagram shows a differential amplifier circuit. It consists of two NPN transistors with their emitters connected to a common emitter node. This node is connected to a 10 kΩ resistor and a component labeled MHC560. The bases of the transistors are connected to a common base node. The collectors are connected to a common collector node. The output is taken from the collector of the left transistor, which is also connected to a component labeled MHC560. The pin number 61 is indicated at the output terminal.</p>
62	IFOUT	 <p>The diagram shows a common-emitter amplifier circuit. It consists of a single NPN transistor. The base is connected to a 20 Ω resistor and a component labeled MHC561. The emitter is connected to ground. The collector is connected to a 100 Ω resistor. The output is taken from the collector, which is also connected to a component labeled MHC561. The pin number 62 is indicated at the output terminal.</p>
63	TAMIF2AGC	 <p>The diagram shows a differential amplifier circuit. It consists of two NPN transistors with their emitters connected to a common emitter node. This node is connected to a 1 kΩ resistor and a component labeled MHC562. The bases of the transistors are connected to a common base node. The collectors are connected to a common collector node. The output is taken from the collector of the right transistor, which is also connected to a component labeled MHC562. The pin number 63 is indicated at the output terminal.</p>
64	TFMIFAGC	 <p>The diagram shows a differential amplifier circuit. It consists of two NPN transistors with their emitters connected to a common emitter node. This node is connected to a diode array and a component labeled MHC563. The bases of the transistors are connected to a common base node. The collectors are connected to a common collector node. The output is taken from the collector of the right transistor, which is also connected to a component labeled MHC563. The pin number 64 is indicated at the output terminal.</p>

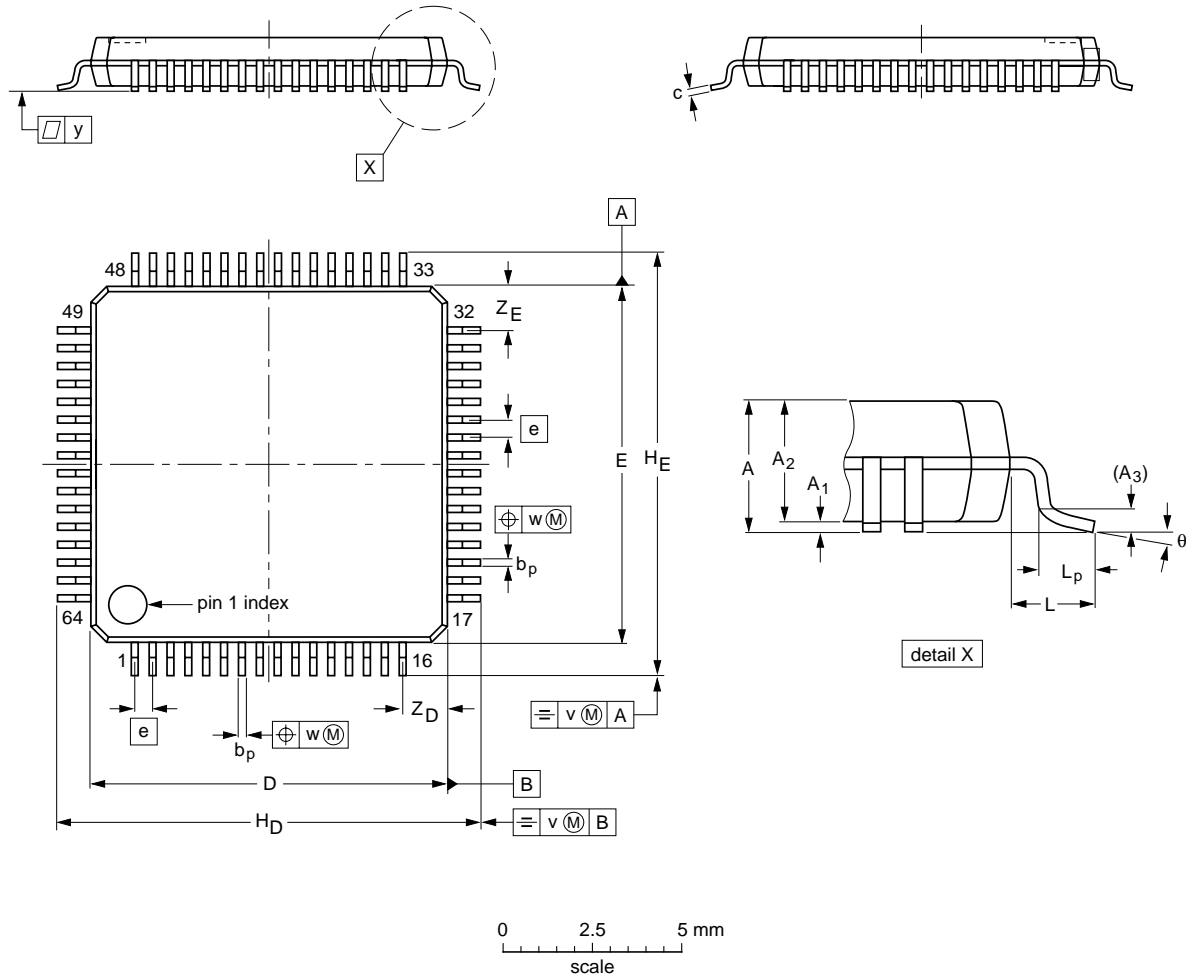
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15 PACKAGE OUTLINE

LQFP64: plastic low profile quad flat package; 64 leads; body 10 x 10 x 1.4 mm

SOT314-2



DIMENSIONS (mm are the original dimensions)

UNIT	A max.	A ₁	A ₂	A ₃	b _p	c	D ⁽¹⁾	E ⁽¹⁾	e	H _D	H _E	L	L _p	v	w	y	Z _D ⁽¹⁾	Z _E ⁽¹⁾	θ
mm	1.6	0.20 0.05	1.45 1.35	0.25	0.27 0.17	0.18 0.12	10.1 9.9	10.1 9.9	0.5	12.15 11.85	12.15 11.85	1	0.75 0.45	0.2	0.12	0.1	1.45 1.05	1.45 1.05	7° 0°

Note

1. Plastic or metal protrusions of 0.25 mm maximum per side are not included.

OUTLINE VERSION	REFERENCES				EUROPEAN PROJECTION	ISSUE DATE
	IEC	JEDEC	JEITA			
SOT314-2	136E10	MS-026				00-01-19 03-02-25

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16 SOLDERING**16.1 Introduction to soldering surface mount packages**

This text gives a very brief insight to a complex technology. A more in-depth account of soldering ICs can be found in our *“Data Handbook IC26; Integrated Circuit Packages”* (document order number 9398 652 90011).

There is no soldering method that is ideal for all surface mount IC packages. Wave soldering can still be used for certain surface mount ICs, but it is not suitable for fine pitch SMDs. In these situations reflow soldering is recommended.

16.2 Reflow soldering

Reflow soldering requires solder paste (a suspension of fine solder particles, flux and binding agent) to be applied to the printed-circuit board by screen printing, stencilling or pressure-syringe dispensing before package placement. Driven by legislation and environmental forces the worldwide use of lead-free solder pastes is increasing.

Several methods exist for reflowing; for example, convection or convection/infrared heating in a conveyor type oven. Throughput times (preheating, soldering and cooling) vary between 100 and 200 seconds depending on heating method.

Typical reflow peak temperatures range from 215 to 270 °C depending on solder paste material. The top-surface temperature of the packages should preferably be kept:

- below 220 °C (SnPb process) or below 245 °C (Pb-free process)
 - for all BGA and SSOP-T packages
 - for packages with a thickness ≥ 2.5 mm
 - for packages with a thickness < 2.5 mm and a volume ≥ 350 mm³ so called thick/large packages.
- below 235 °C (SnPb process) or below 260 °C (Pb-free process) for packages with a thickness < 2.5 mm and a volume < 350 mm³ so called small/thin packages.

Moisture sensitivity precautions, as indicated on packing, must be respected at all times.

16.3 Wave soldering

Conventional single wave soldering is not recommended for surface mount devices (SMDs) or printed-circuit boards with a high component density, as solder bridging and non-wetting can present major problems.

To overcome these problems the double-wave soldering method was specifically developed.

If wave soldering is used the following conditions must be observed for optimal results:

- Use a double-wave soldering method comprising a turbulent wave with high upward pressure followed by a smooth laminar wave.
- For packages with leads on two sides and a pitch (e):
 - larger than or equal to 1.27 mm, the footprint longitudinal axis is **preferred** to be parallel to the transport direction of the printed-circuit board;
 - smaller than 1.27 mm, the footprint longitudinal axis **must** be parallel to the transport direction of the printed-circuit board.

The footprint must incorporate solder thieves at the downstream end.

- For packages with leads on four sides, the footprint must be placed at a 45° angle to the transport direction of the printed-circuit board. The footprint must incorporate solder thieves downstream and at the side corners.

During placement and before soldering, the package must be fixed with a droplet of adhesive. The adhesive can be applied by screen printing, pin transfer or syringe dispensing. The package can be soldered after the adhesive is cured.

Typical dwell time of the leads in the wave ranges from 3 to 4 seconds at 250 °C or 265 °C, depending on solder material applied, SnPb or Pb-free respectively.

A mildly-activated flux will eliminate the need for removal of corrosive residues in most applications.

16.4 Manual soldering

Fix the component by first soldering two diagonally-opposite end leads. Use a low voltage (24 V or less) soldering iron applied to the flat part of the lead. Contact time must be limited to 10 seconds at up to 300 °C.

When using a dedicated tool, all other leads can be soldered in one operation within 2 to 5 seconds between 270 and 320 °C.

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16.5 Suitability of surface mount IC packages for wave and reflow soldering methods

PACKAGE ⁽¹⁾	SOLDERING METHOD	
	WAVE	REFLOW ⁽²⁾
BGA, LBGA, LFBGA, SQFP, SSOP-T ⁽³⁾ , TFBGA, VFBGA DHVQFN, HBCC, HBGA, HLQFP, HSQFP, HSOP, HTQFP, HTSSOP, HVQFN, HVSON, SMS PLCC ⁽⁵⁾ , SO, SOJ LQFP, QFP, TQFP SSOP, TSSOP, VSO, VSSOP PMFP ⁽⁸⁾	not suitable not suitable ⁽⁴⁾ suitable not recommended ⁽⁵⁾⁽⁶⁾ not recommended ⁽⁷⁾ not suitable	suitable suitable suitable suitable suitable not suitable

Notes

- For more detailed information on the BGA packages refer to the “(LF)BGA Application Note” (AN01026); order a copy from your Philips Semiconductors sales office.
- All surface mount (SMD) packages are moisture sensitive. Depending upon the moisture content, the maximum temperature (with respect to time) and body size of the package, there is a risk that internal or external package cracks may occur due to vaporization of the moisture in them (the so called popcorn effect). For details, refer to the Drypack information in the “Data Handbook IC26; Integrated Circuit Packages; Section: Packing Methods”.
- These transparent plastic packages are extremely sensitive to reflow soldering conditions and must on no account be processed through more than one soldering cycle or subjected to infrared reflow soldering with peak temperature exceeding $217\text{ °C} \pm 10\text{ °C}$ measured in the atmosphere of the reflow oven. The package body peak temperature must be kept as low as possible.
- These packages are not suitable for wave soldering. On versions with the heatsink on the bottom side, the solder cannot penetrate between the printed-circuit board and the heatsink. On versions with the heatsink on the top side, the solder might be deposited on the heatsink surface.
- If wave soldering is considered, then the package must be placed at a 45° angle to the solder wave direction. The package footprint must incorporate solder thieves downstream and at the side corners.
- Wave soldering is suitable for LQFP, TQFP and QFP packages with a pitch (e) larger than 0.8 mm; it is definitely not suitable for packages with a pitch (e) equal to or smaller than 0.65 mm.
- Wave soldering is suitable for SSOP, TSSOP, VSO and VSSOP packages with a pitch (e) equal to or larger than 0.65 mm; it is definitely not suitable for packages with a pitch (e) equal to or smaller than 0.5 mm.
- Hot bar or manual soldering is suitable for PMFP packages.

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17 DATA SHEET STATUS

LEVEL	DATA SHEET STATUS ⁽¹⁾	PRODUCT STATUS ⁽²⁾⁽³⁾	DEFINITION
I	Objective data	Development	This data sheet contains data from the objective specification for product development. Philips Semiconductors reserves the right to change the specification in any manner without notice.
II	Preliminary data	Qualification	This data sheet contains data from the preliminary specification. Supplementary data will be published at a later date. Philips Semiconductors reserves the right to change the specification without notice, in order to improve the design and supply the best possible product.
III	Product data	Production	This data sheet contains data from the product specification. Philips Semiconductors reserves the right to make changes at any time in order to improve the design, manufacturing and supply. Relevant changes will be communicated via a Customer Product/Process Change Notification (CPCN).

Notes

1. Please consult the most recently issued data sheet before initiating or completing a design.
2. The product status of the device(s) described in this data sheet may have changed since this data sheet was published. The latest information is available on the Internet at URL <http://www.semiconductors.philips.com>.
3. For data sheets describing multiple type numbers, the highest-level product status determines the data sheet status.

18 DEFINITIONS

Short-form specification — The data in a short-form specification is extracted from a full data sheet with the same type number and title. For detailed information see the relevant data sheet or data handbook.

Limiting values definition — Limiting values given are in accordance with the Absolute Maximum Rating System (IEC 60134). Stress above one or more of the limiting values may cause permanent damage to the device.

These are stress ratings only and operation of the device at these or at any other conditions above those given in the Characteristics sections of the specification is not implied. Exposure to limiting values for extended periods may affect device reliability.

Application information — Applications that are described herein for any of these products are for illustrative purposes only. Philips Semiconductors make no representation or warranty that such applications will be suitable for the specified use without further testing or modification.

19 DISCLAIMERS

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20 PURCHASE OF PHILIPS I²C COMPONENTS

Purchase of Philips I²C components conveys a license under the Philips' I²C patent to use the components in the I²C system provided the system conforms to the I²C specification defined by Philips. This specification can be ordered using the code 9398 393 40011.

Philips Semiconductors – a worldwide company

Contact information

For additional information please visit <http://www.semiconductors.philips.com>. Fax: +31 40 27 24825

For sales offices addresses send e-mail to: sales.addresses@www.semiconductors.philips.com.

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Printed in The Netherlands

753503/01/pp56

Date of release: 2003 Oct 21

Document order number: 9397 750 11378

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